

بِسْمِ اللَّهِ الرَّحْمَنِ الرَّحِيمِ

الحلول المختارة لطلاب الهندسة والعمارة

القياسية الكهربائية والإلكترونية

أعداد

محمد بن عبد الله العاصمي

الأستاذ المساعد بقسم الهندسة الكهربائية والإلكترونية

جامعة أم القرى

## بِسْمِ اللَّهِ الرَّحْمَنِ الرَّحِيمِ

الحمد لله والصلوة والسلام على سيدنا رسول الله وعلى آله وصحبه وسر والاه  
وبعد، فإذنا مجتمعة من المسائل والحلول في مادة القياس الكلاسيكية  
والالكتونية بدأنا في د/ علي شادي ود/ محمد الهناوي ود/ عبد الحميد  
يوسف من أعضاء هيئة التدريس بقسم الهندسة الكلاسيكية بجامعة  
الملك عبد العزيز بإبارة تعاقبهم على تدريس علم بضع سنوات.  
وقد تمت بمراجعتكم وتبويركم والزيادة عليه بما يغطي الفروع المنهجية  
بين ما يدرسه القسم المذكور بعاليه وما يدرسه بقسم الهندسة  
الكلاسيكية والكمبيوترية بجامعة أم القرى حيث نحني الأول بقياسي مختلف  
الكميات الفيزيائية ونحني الآخر بقياسي الكميات الكلاسيكية من خصائصها.  
وقد أدرجت المسائل أوائل أو أتبعت بالحلول لبعض من حسب  
التبويب الذي حرت عليه بإبارة تدريس هذه المادة بقسم الهندسة  
الكلاسيكية والكمبيوترية بجامعة أم القرى،

والله شال أنه يجعل هذا العمل مفيداً وأنه يجزي من شارك  
فيه خير جزاء.

محمد الهناوي  
١٤١٧/١١/٢٦

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## PROBLEM SET *One, Uncertainty.*

1-1

A capacitive transducer of two parallel plates of overlapping area of  $5 \text{ cm}^2 \pm 0.5\%$  immersed in water. The capacitance,  $c$ , was found to be 950 PF. Calculate the Separation,  $d$  between the two plates and the sensitivity  $(\partial c / \partial d)$  of this transducer.

If the allowed uncertainty in  $d$  is 3%, what is the maximum permissible uncertainty in  $C$ ?

Physical Constants:

Magnetic Permeability for free space ( $\mu_0$ ) is  $4\pi \times 10^{-7} \text{ H/m}$

Permittivity of Vacuum ( $\epsilon_0$ ) is  $8.854 \text{ PF/m}$

Relative permittivity for water ( $\epsilon_r$ ) is 81

1-2

The thermistor is a **semiconductor** resistance transducer used in the measurement of temperature. The output-input relation for that transducer is

$$R = R_0 \exp\left[\beta_0 \left(\frac{1}{T} - \frac{1}{T_0}\right)\right]$$

where  $R_0$  is the value of resistance at a reference temperature  $T_0$ . At the unknown temperature  $T$ , the resistance  $R$  is measured, then  $T$  is calculated as:

$$T = \frac{T_0}{1 + (T_0/\beta_0) \ln(R/R_0)}$$

For  $T_0 = 300^\circ\text{K}$ ,  $\beta_0 = 3420^\circ\text{K}$ ,  $R_0 = 1\text{K}\Omega \pm 30$ ,  $R = 2\text{K}\Omega \pm 2\%$

- a) Calculate the nominal value of  $T$ .
- b) Find the sensitivity  $(\partial R / \partial T)$  of the transducer at the given operating point.

c) Show that:  $(\Delta T / T)^2 = (T / \beta_0)^2 [(\Delta R_0 / R_0)^2 + (\Delta R / R)^2]$

1-3

A voltage source of an internal resistance  $R_0 = 0.5 \text{ k}\Omega$  is measured 8 times with a moving coil meter. The following readings were obtained.

6.73, 6.74, 6.69, 6.71, 6.70, 6.73, 6.72, 6.74 volts

The same source was also measured by a recently calibrated digital voltmeter whose input resistance is no less than 20 MO. The following 6 readings were obtained:

7.073, 7.069, 7.072, 7.074, 7.071, 7.073 volts.

- a) Find the precision and accuracy of both sets of readings
- b) How should the readings of the moving coil meter be recalibrated?

1-4

A spring scale is a transducer that converts weight,  $W$ , to deflection,  $Y$ .



Its output-input relation is given by

$$Y = \frac{8 D^3 N}{E X^4} \cdot W$$

where  $D$  = mean coil diameter of the spring,  $X$  = its steel wire diameter,  $E$  = torsional elastic modulus for steel, and  $N$  = number of coil turns.

The deflection was measured to be  $Y = (10 \pm 0.02)$  cm. for the following transducer specifications:

$$D = 2.5 \text{ cm.} \pm 2\%, \quad X = 2.5 \text{ mm.} \pm 1\%$$

$$E = 80 \times 10^9 \text{ Pa} \pm 4\%, \quad \text{and } N = 50 \pm 0.5$$

- Calculate the nominal value of  $W$ .
- Calculate the uncertainty in  $W$ . Rank the various variables according to their contribution to  $\Delta W$ .
- Find the nominal value of the transducer's static sensitivity. Plot this value as a function of the input  $W$ .

1-5

A satellite is observed so that its speed may be determined. On the first observation, it was found at a distance  $R$  from the observer =  $(30,000 \pm 10)$  km. Five seconds later, this distance has increased by  $r = (125.0 \pm 0.5)$  km and the change in angle was  $\theta = (0.00750 \pm 0.00002)$  radians. What is the speed of the satellite, assuming that it moves in a straight line and with constant speed in the given direction ?

(Hint = Use the cosine formula:

$$d^2 = (R + r)^2 + R^2 - 2R(R + r) \cos \theta$$

$$= r^2 + 2R(R + r)(1 - \cos \theta)$$

and note that  $1 - \cos \theta = \theta^2/2 - \theta^4/24 + \dots$ ).

1-6

A strain whose true value is conventionally known as  $340 \mu\text{m/m}$  was measured by a certain gauge. The following 12 readings were obtained:

346, 345, 347, 339, 342, 345, 347, 343, 340, 344, 351, 345  $\mu\text{m/m}$ .

calculate:

- the average value and the bias of this set of readings.
- the precision and accuracy of the gauge.

1-7

The output expression for a multiplier is

$$Z = y_1 y_2$$

However, to include all sources of error, this expression is rewritten as:

$$Z = P \cdot y_1^{I_1} \cdot y_2^{I_2} + S$$

where  $y_1 = y_{10} \pm W_{10}$ ,  $y_2 = y_{20} \pm W_{20}$ ,  $P = 1 \pm W_0$

$$I_1 = 1 \pm W_{11}, I_2 = 1 \pm W_{21}, S = 0 \pm W_3$$

Estimate an upper bound on the error in Z

1-8

A metallic resistance thermometer has a linear variation of resistance with temperature  $R = R_0 [1 + \alpha (T - T_0)]$ .

The resistance  $R_0 = 20 \text{ KO} \pm 0.1\%$ , while at a temperature T the resistance R is found to be  $R = 30 \text{ KO} \pm 0.1\%$ . The coefficient  $\alpha$  is  $0.00392 \text{ } ^\circ\text{K}^{-1}$ .

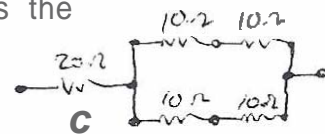
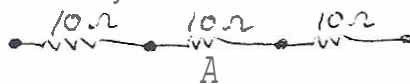
- Write an explicit expression for T.
- calculate the nominal value of T.
- Find the static sensitivity ( $\partial R / \partial T$ ) of the thermometer.
- Show that the uncertainty  $\Delta T$  in T is given by

$$(\Delta T)^2 = (\Delta T_0)^2 + (1/\alpha^2) (R/R_0)^2 [(\Delta R_0/R_0)^2 + (\Delta R/R)^2]$$

e) Find the value of uncertainty in T.

1-9

five resistors are available, one of  $20 \Omega$  and four of  $10 \Omega$  each. The uncertainty of the  $20 \Omega$  resistor is 5% and that of each  $10 \Omega$  resistor is 10%. Three possible connections using these resistors are shown below. Which connection would you use to obtain a resistance of  $30 \Omega$  with the least uncertainty? What is the uncertainty of this best connection?



1-10

The electronic counter can be used for measuring the time period of periodic signals. Show that the uncertainty in the measurement can be reduced by a factor of  $1/\sqrt{N}$  if the average of N time periods is taken.

Hint:  $T_{av} = 1/N (T_1 + T_2 + \dots + T_N)$ .

The  $T_i$ 's are statistically independent,  $T_i = T + \Delta T$

1-11

The discharge coefficient  $C$  of an orifice can be found by collecting the water that flows through during a timed interval when it is under a constant head  $h$ . The formula is

$$C = \frac{W}{t \rho A \sqrt{2gh}}$$

Find  $C$  and its uncertainty if

$$W = (865 \pm 0.5) \text{ lb}_m, A = \frac{\pi d^2}{4}, d = (0.500 \pm 0.001)''$$

$$t = (600.0 \pm 2) \text{ s}, g = 32.17 \text{ ft/s}^2 \pm 0.1\%$$

$$\rho = 62.36 \text{ lb}_m/\text{ft}^3 \pm 0.1\%$$

$$h = (12.02 \pm 0.01) \text{ ft}$$

1-12

The dc current in a resistor  $R = 10 \text{ k}\Omega \pm 5\%$  is measured to be  $I = 10 \text{ mA} \pm 1\%$ .  
Find the power it dissipates.

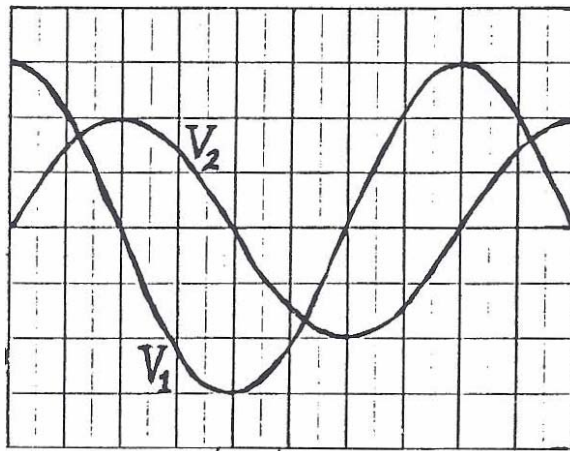


## Problem Set Two, Scope.

2-1

Two sinusoidal signals  $V_1$  and  $V_2$  are applied to an oscilloscope in dual trace operation ( $V_1$  applied to CH.1 and  $V_2$  applied to CH.2). The trigger source is CH.1 and the various sensitivities are  $S_{CH1} = 20$  mV/cm,  $S_{CH2} = 1$  V/cm,  $S_x = 20$  ms/cm. For the dual trace shown, find

- the peak to peak values of  $V_1$  and  $V_2$
- the ratio of  $V_2$  to  $V_1$  in decibels.
- the time period and the frequency of both signals.
- the trigger level and trigger slope.
- the phase shift between  $V_1$  and  $V_2$ . Does  $V_1$  lead or lag  $V_2$ ?



2-2

The two voltage signals

$$V_1(t) = 40 \sin(628t + \pi/5) \text{ V}$$

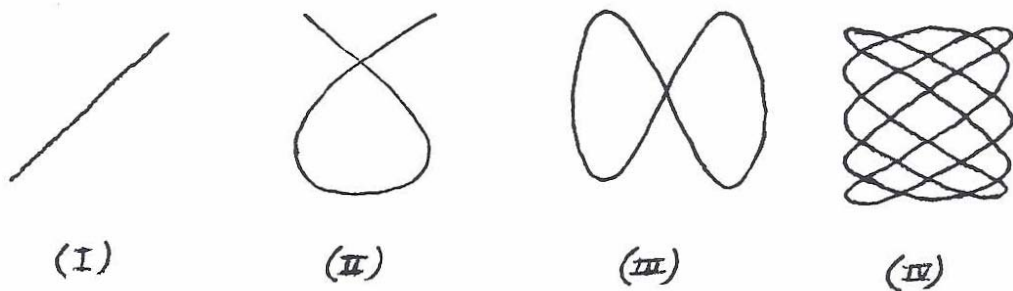
$$V_2(t) = 60 \sin(628t) \text{ V}$$

are applied to channel 1 and 2 respectively of a dual-trace oscilloscope. Sketch the trace on the 8 cm x 10 cm CRT screen, given that:

- The vertical sensitivity of both channels = 20 V/cm.
- The time base sensitivity = 2 ms/cm
- The trigger source is channel 2, with a trigger level of 30 V and trigger slope = positive.

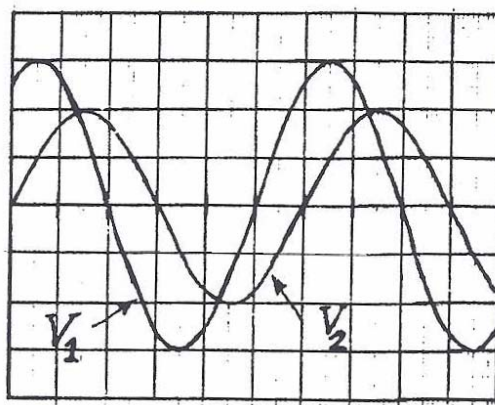
5

- ⑥ If the oscilloscope is switched to XY operation, sketch the pattern that results on the screen if  $V_1(t)$  is applied to the X-input and  $V_2(t)$  is applied to the Y-input and both horizontal and vertical sensitivities are kept at 20 V/cm.
- ⑦ For a horizontal frequency  $f_H = 1 \text{ kHz}$ , the following Lissajous figures were obtained. Find the corresponding vertical frequency for each figure.



2-3

Two sinusoidal voltages  $V_1(t) = V_1 \sin(\omega t + \theta_1)$  and  $V_2(t) = V_2 \sin(\omega t + \theta_2)$  are applied to CH.1 and CH.2 respectively of an oscilloscope. The two vertical sensitivities are both equal to 50 mV/cm, and the time-base sensitivity is 20 ms/cm. For the shown trace determine the amplitudes  $V_1$  and  $V_2$ , the circular frequency  $\omega$ , and the phase shift  $\theta = \theta_2 - \theta_1$ .



2-4

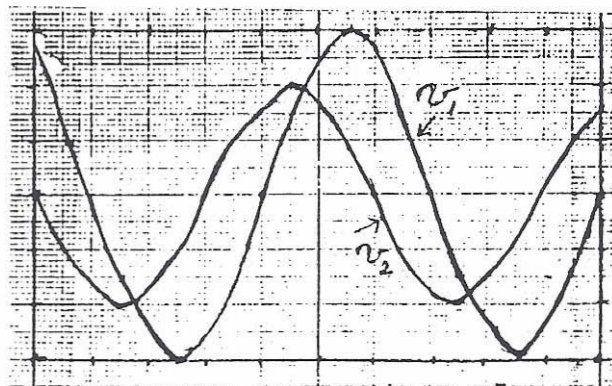
Two sinusoidal voltages  $V_1$  and  $V_2$  are applied to an oscilloscope,  $V_1$  applied to channel 1 and  $V_2$  applied to channel 2.

Given:

- Channel 1 Vertical Sensitivity = 0.5 V/cm.
- Channel 2 Vertical Sensitivity = 1.0 V/cm.
- Time Base Sensitivity = 10 m Sec/cm
- Trigger Source is Channel 1.
- See Fig. below

Determine:

- Peak to peak value of  $V_2$
- Frequency of both signals.
- Phase-shift between  $V_2$  and  $V_1$ . Which is leading?
- Trigger level.
- Trigger slope, whether + ve or - ve.



2-5

An oscilloscope was used for the measurement of phase shift  $\theta$  between two signals  $V_1$  and  $V_2$  of the same frequency. The following results were obtained:

- for the ellipse method [ $\theta = \sin^{-1} (y_0/y_m)$ ]

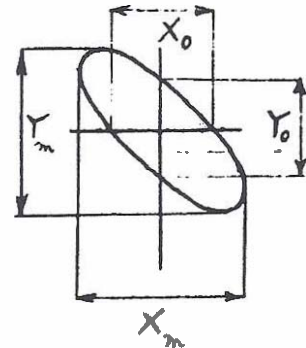
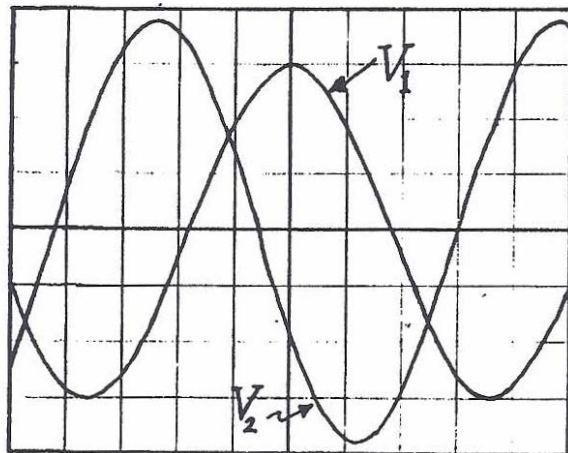
$$y_0 = (3.5 \pm 0.05) \text{ cm}, y_m = (5 \pm 0.05) \text{ cm}$$

- for the dual-trace method [ $\phi = (d/D) 2\pi$  rad]

$$d = (1 \pm 0.05) \text{ cm}, \quad D = (8 \pm 0.05) \text{ cm}.$$

In both cases determine the phase shift  $\phi$  and its uncertainty. Can any of the methods (a) or (b) be used to determine if  $V_1$  leads or lags  $V_2$ ?

2-6



Two sinusoidal voltages  $V_1$  and  $V_2$  are applied to an oscilloscope in dual trace operation ( $V_1$  applied to channel 1 and  $V_2$  applied to channel 2). The vertical sensitivities are  $S_{CH1} = 10 \text{ mV/cm}$  and  $S_{CH2} = 0.05 \text{ V/cm}$ . The time base has  $S_H = 10 \text{ mS/cm}$ . The trigger source is CH.2. For the dual trace shown, find:

- The peak to peak values of  $V_1$  and  $V_2$ .
- The time period and frequency of both signals.
- The trigger level and trigger slope.
- The phase shift between  $V_1$  and  $V_2$ . Does  $V_1$  lead or lag  $V_2$ ?

Now, the oscilloscope is switched to XY operation  $V_1$  is connected, to the X input with  $S_X = 10 \text{ mV/cm}$ , and  $V_2$  is connected to the Y input with  $S_Y = 0.5 \text{ V/cm}$ , so that the above ellipse results. Find

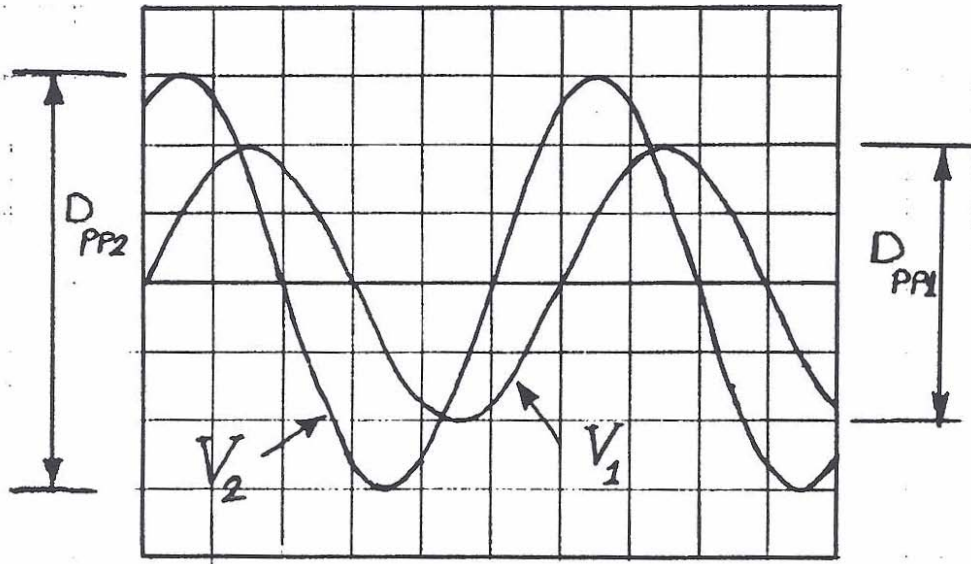
- The distances  $Y_m$ ,  $Y_0$ ,  $X_m$ , and  $X_0$ .

2-7

The input and output to an amplifier are the two sinusoidal voltages *shown*  $V_1$ , and  $V_2$  respectively. These two voltages are applied to an oscilloscope in dual trace operation ( $V_1$  applied to CH.1 and  $V_2$  applied to CH.2)

CH.1 vertical sensitivity :  $S_1 = 20 \text{ mV/Cm}$   
 CH.2 vertical sensitivity :  $S_2 = 0.5 \text{ V/Cm}$   
 Time base SH =  $10 \text{ ms/Cm}$   
 Trigger source is CH.2

The trace obtained is **as** shown in the figure below. Assume an uncertainty of 0.5 mm. in all distances measured.



— if the gain  $G$  of the amplifier is defined in dB by:

$$G = 20 \log_{10} \left( \frac{S_2 \cdot D_{pp2}}{S_1 \cdot D_{pp1}} \right)$$

Where  $S_1, S_2, D_{pp1}$  and  $D_{pp2}$  are defined as above. Show that the uncertainty in the gain *is* given by:

$$(W_G)^2 = (20 \log_{10} e)^2 \left[ \left( \frac{W_{D_{pp2}}}{D_{pp2}} \right)^2 + \left( \frac{W}{D_{pp1}} \right)^2 \right]$$

- Calculate the value of  $G$  and  $W_G$  for the above case.
- Find the phase shift between  $V_1$  and  $V_2$  and its uncertainty. Does  $V_1$  lead or lag  $V_2$ ?

2-8

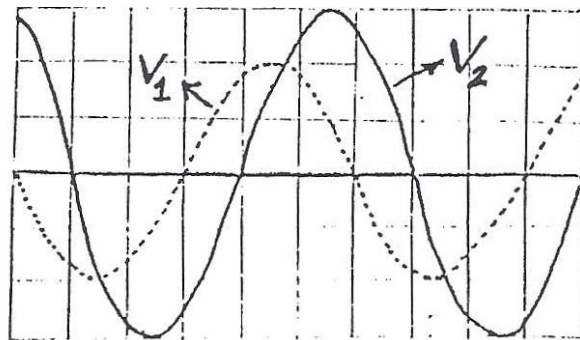
Two sinusoidal voltages  $V_1$  and  $V_2$  are applied to the oscilloscope;  $V_1$  applied to CH.1 and  $V_2$  applied to CH.2

- CH.1 vertical sensitivity = 50 mV/cm
- CH.2 vertical sensitivity = 50 mV/cm
- Time base = 20 ns/cm
- Trigger source = CH.1

For the shown trace determine:

- Peak to peak value of  $V_2$
- Time period of both signals
- Phase shift
- Trigger level
- Trigger slope
- Assuming an uncertainty of  $\pm 0.5$  mm in all distances measured, what is the percentage uncertainties in the results of the first three parts

If the 2 signals are applied to the oscilloscope in XY operation, sketch the resulting display.



\_\_\_\_\_



2-9

Two voltages  $e_1$  and  $e_2$  are represented by

$$e_1 = 100 \sin(314 t) \text{ volt}$$

$$e_2 = -150 \sin(628 t) \text{ volt}$$

These are applied to channel 1 ( $Y_1$ ) and channel 2 ( $Y_2$ ) of a double beam oscilloscope respectively

a) Sketch the trace on the CRT screen, if

Vertical sensitivity of both channels = 50 V/cm.

Time Base = 2 m Sec/cm

Width of the screen = 10 cm

Height of the screen = 8 cm

Trigger Source-Channel 1

Trigger Level = 0 (zero)

Trigger slope = positive

b) If the time base is switched off and XY operation is used with  $Y_2$  input applied to Y-deflecting plates and  $Y_1$  input applied to X deflecting plates, sketch graphically the pattern on the CRT screen.

2-10

Two voltages  $V_1$  and  $V_2$  were applied to dual channel oscilloscope.  $V_1$  applied to Ch.1 and  $V_2$  applied to Ch.2.

Ch.1 Vertical Sensitivity = 50 mv/cm.

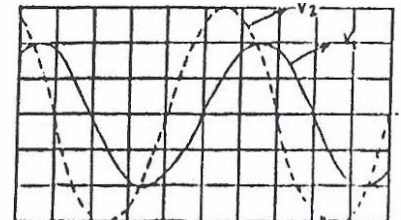
Ch.2 Vertical Sensitivity = 0.5v/cm.

Time base = 0.1 m sec./cm.

Trigger source Ch.1

For the shown scope trace, determine.

- Peak to peak value of  $V_1$ ,
- Peak to peak value of  $V_2$
- frequency of both signals
- Phase shift  $V_2$  (leads)  $V_1$  by 7
- Trigger level
- Slope (+ve or -ve)



Sketch the scope waveform for a sine wave input 30 mv peak at a frequency 1000 Hz, given that

- Vertical sensitivity 10 mv/cm
- Time sensitivity 0.2 m sec/cm.
- Screen width 10 cm.
- Trigger level 0v
- Slope:- negative.

2-12

In problem 2 above sketch the wave form of another signal 20 mv peak at the same frequency and having a phase shift  $50^\circ$  leading, given that the second channel sensitivity is 10 mv/cm.

2-13

Sketch the resulting pattern on the Oscilloscope, vertical scale = 0.5 v/cm, horizontal scale 0.5 v/cm, XY Mode

- |                                 |                                   |
|---------------------------------|-----------------------------------|
| a) $V_x = 1 \sin wt$            | $V_y = 45 \sin (wt + 60^\circ)$   |
| b) $V_x = 2 \sin wt$            | $V_y = 1 \sin (wt - 60^\circ)$    |
| c) $V_x = 0.5 \sin wt$          | $V_y = 1.5 \sin (wt + 120^\circ)$ |
| d) $V_x = 1.5 \sin wt$          | $V_y = 2 \sin (wt - 120^\circ)$   |
| e) $V_x = 2 \sin wt$            | $V_y = 1 \cos wt$                 |
| f) $V_x = 2 \sin wt$            | $V_y = 2 \cos wt$                 |
| g) $V_x = 2 \sin (wt + \theta)$ | $V_y = 1.5 \sin (wt + \theta)$    |
| h) $V_x = 2 \sin (wt + \theta)$ | $V_y = 1.5 \sin (wt - \theta)$    |

2-14 For the Lissajous

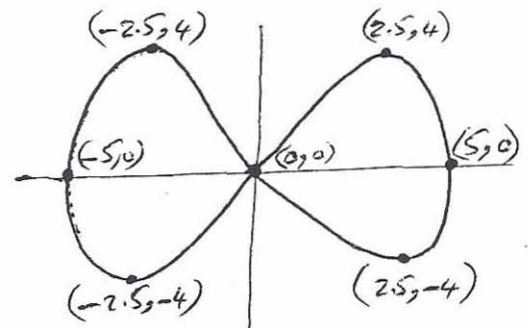
figure shown:

X-Sensitivity = 5V/cm

$V_y(t) = 16 \sin(wt + \phi)$  Volts

Find:

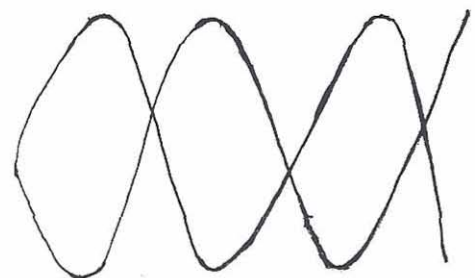
- Y-Sensitivity
- $V_x(t)$
- Sketch trace on sweep mode given that  $\omega = 40\pi \text{ rad/ms}$  time base 5  $\mu\text{s/cm}$ , Trig. from ch X, level 2 cm & slope positive



2-15

A scope is put into XY mode and the frequency of the X-Input was 50 Hz.

For the figure shown what is the Y-Input frequency?



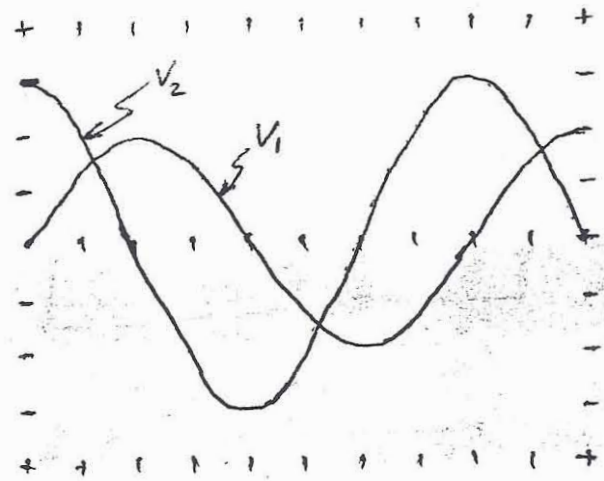
**2-16** If  $V_x$  was given by  $V_x = 2 \sin 100\pi t$ , Volts  $\neq$   
 $V_y = 4 \cos 100\pi t$ , Volts,  
 $S_x = 1V/Cm$   $\neq$   $S_y = 2V/Cm$

- a) Plot screen on XY mode  
 b) Plot screen on time-sweep mode given that:  
 -  $V_x$  is applied to ch 1  $\neq$   $V_y$  to ch 2, with same sensitivities,  
 - Trig. source is ch. 2, positive slope with level of 2 Volts  
 - Time-base is  
 I)  $1\mu\text{sec}/Cm$   $\neq$  II)  $2\text{msec}/Cm$

**2-17** For the shown scope-Screen,

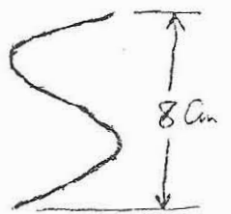
given:

- $V_1$  applied to ch 1 at  $1V/Cm$
- $V_2$  applied to ch 2 at  $20mV/Cm$
- Trig. source is ch 1
- Time base at  $20\text{msec}/Cm$
- Uncertainties of  $1\text{mm}$



- Find
- a-  $V_{1,rms}$ ,  $V_{2,rms}$  with uncertainties
  - b- frequency of both signals with uncertainties
  - c- trig. level  $\neq$  slope
  - d- phase shift in degrees
  - e- screen at XY mode operation.

**2-18** An xy mode screen shows the figure showing, what is the trace on sweep mode if  $X = 2 \sin 60\pi t Cm$ ,  
 Trig. level  $0 Cm$ , slope +ve, source ch x,  
 Sweep rate =  $10\mu\text{sec}/Cm$ .



**2-19** Two signals  $V_1$  &  $V_2$  are applied to ch1 & ch2 respectively of a scope. They are given by:

$$V_1(t) = 15 \sin 50\pi t \quad \text{volts } f$$

$$V_2(t) = 20 \sin 100\pi t \quad \text{volts. Sketch the trace for:}$$

a) Trigger Mode, given:  $S_1 = 5 \text{ V/cm}$ ,  $S_2 = 10 \text{ V/cm}$ ,

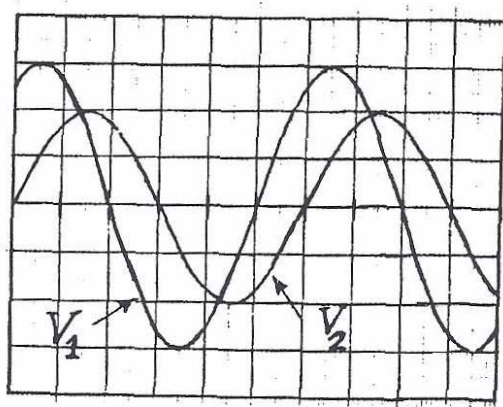
Trigger source ch1, slope +ve, level zero &

Time base of  $2 \text{ msec/cm}$

b) XY-Mode, given the above sensitivities, with ch1 as x & ch2 as y.

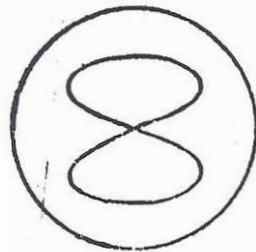
**2-20** Two sinusoidal voltages  $v_1(t) = V_1 \sin(\omega t + \phi_1)$  and  $v_2(t) = V_2 \sin(\omega t + \phi_2)$  are applied to CH.1 and CH.2 respectively of an oscilloscope. The two vertical sensitivities are

both equal to  $50 \text{ mV/cm}$ , and the time-base sensitivity is  $20 \text{ ms/cm}$ . For the shown trace determine the amplitudes  $V_1$  and  $V_2$ , the circular frequency  $\omega$ , and the phase shift  $\phi = \phi_2 - \phi_1$ .

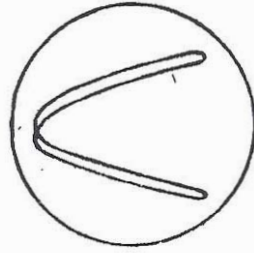


2-21

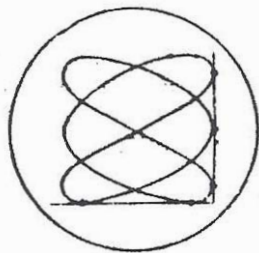
For the shown Lissajous figures  $f_x = 240 \text{ Hz}$ .  
Find  $f_y$ .



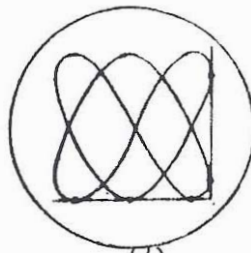
(a)



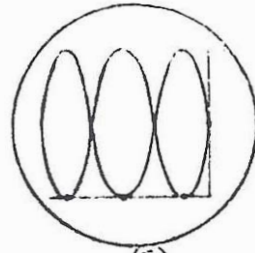
(b)



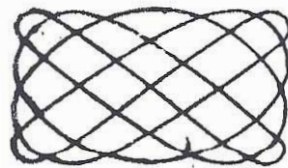
(c)



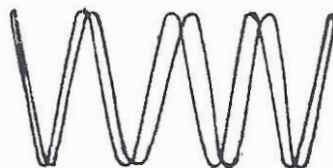
(d)



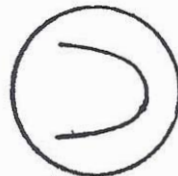
(e)



(f)



(g)



(h)



(i)





## Problem Set Three, Counters.

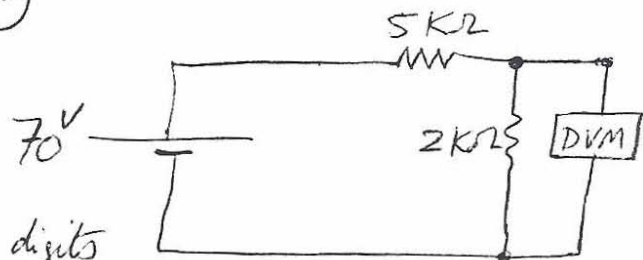
3-1 A DFM set to measure the period of a signal reads  $7125.3925 \mu\text{s}$ . Find the interval in which the correct period falls. Find also the frequency and its uncertainty.

---

3-2 The DFM was now set to measure the frequency of the above signal, what do you think the reading would be on the KHz scale range. Which method do you prefer to use for this signal? Why?

---

3-3 A DVM was used to determine the voltage of the following circuit. If the sensitivity of the DVM was  $10 \text{K}\Omega/\text{V}$  of FSD and it has three ranges  $1000 \text{V}$ ,  $100 \text{V}$  &  $10 \text{V}$ , with  $4\frac{1}{2}$  digits what range would you select? and what is the error in this-range reading. Explain your comments.





3-4 A 4 digit DFM has the following ranges for unit count: 1ms, 10ms, 100ms, 1s, 10s, .1 Hz, 1 Hz, 10 Hz, 100 Hz & 1 KHz.

- If the reading on 100 Hz range was 0095, what is the likely frequency? Can you measure it better? How?
- If the frequency was 2.745 KHz, what will the display be for each range?

3-5 A DFC screen can have full illumination of

1 0 0 0 0 0 0 0  
1 0 0 0 0 0 0 0

with unit count range of:

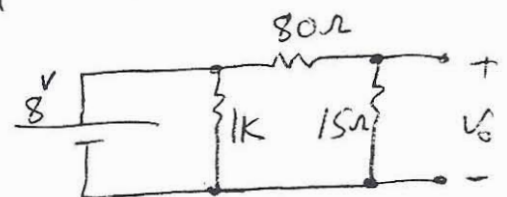
1ms, 10ms, 100ms, 1s, 10s, 10 Hz, 100 Hz, 1 KHz, 10 KHz & 1 MHz

- What is the no. of digits?
- What is the best reading for  $f = \pi$  Hz?
- On what range is it obtained?
- What is % uncertainty?
- What is the range of measurement of  $f$  using this DFC?

3-6 The above DFC was made DVM with sensitivity of  $1 \text{ K}\Omega/\text{V}$  of FSD. The FSD ranges are

2mV, 0.2V, 20V & 2KV

Find: (a) best range,  
(b) Reading &  
(c) % error,



for measuring  $V_0$  shown in the figure.

**3-7** A  $5\frac{1}{2}$  Digits FC was used to measure 60 Hz signal.

If LSD can range as follows 10 ms, 1 s, 100 s, 100 Hz, 10 Hz  
& 1 Hz, what is the best reading. Find its  $\%e$

---

## PROBLEM SET FOUR, *Moving Coil Meter.*

A moving coil instrument has the following data:

No. of turns	=	100
Width of coil	=	2 cm
Depth of coil	=	3 cm
Flux density in the gap	=	0.1 Wb/m <sup>2</sup>

Calculate the deflecting torque when carrying a current of 10 mA. Also calculate the deflection if the control spring constant is  $20 \times 10^{-7}$  N-m per degree.

---

The coil of a mc voltmeter is 4 cm long and 3 cm wide and has 100 turns on it. The control spring exerts a torque of  $2.4 \times 10^{-4}$  N-m when the deflection is 100 divisions on full scale. If the flux density of the magnetic field in the air-gap is 0.1 Wb/m<sup>2</sup>, estimate the resistance that must be put in series with the coil to give one volt per division. The resistance of the voltmeter coil may be neglected.

---

**4-3** Design a multi-range d.c. ammeter using a basic movement with an internal resistance  $R = 50 \Omega$  and a full scale deflection current  $I = 1 \text{ mA}$ . The ranges required are 0-10 mA, 0-50 mA, 0-100 mA and 0-500 mA

---

**4-4** A mc instrument gives full scale deflection of 10 mA when the P.D. across its terminals is 100 mV. Calculate (a) the shunt resistance for a full scale deflection corresponding to 100 A and (b) the resistance for full scale reading with 1000 V. Calculate the power dissipated in each case.

---

**4-5** A basic d'Arsonval meter movement with an internal resistance  $R_m = 100 \Omega$ , a full scale current of  $I = 1 \text{ mA}$ , is to be converted into a multi-range d.c. voltmeter with ranges of 0-10V, 0-50V; 0-250V and 0-500V. Find the values of various resistance using the potential divider arrangement.

---

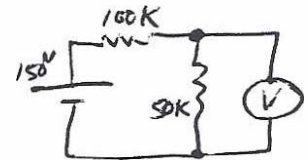
**4-6** Calculate the multiplying power of a shunt of  $200 \Omega$  resistance used with a galvanometer of  $1000 \Omega$  resistance. Determine the value of shunt resistance to give a multiplying power of 50.

---

4-7

It is desired to measure the voltage across a  $50\text{ K}\Omega$  resistor in the circuit shown in Fig. Two voltmeters are available for this purpose.

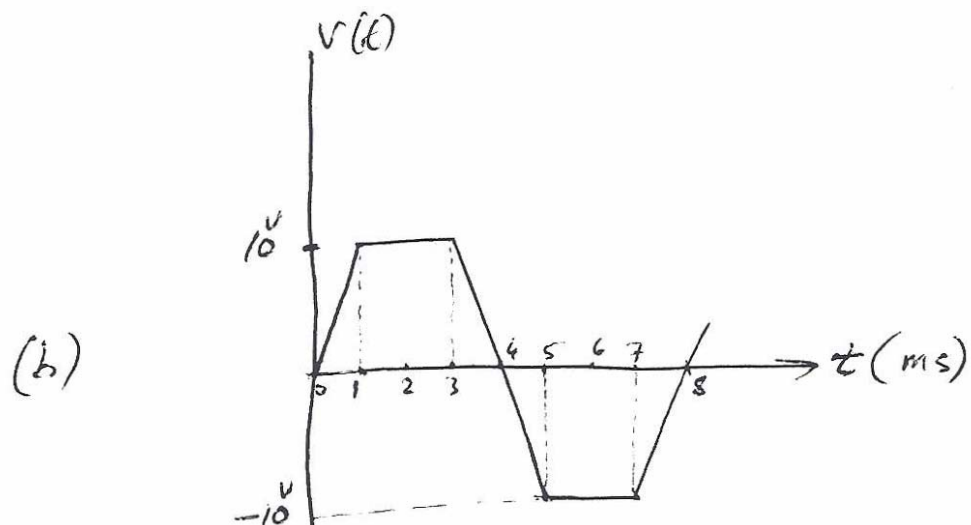
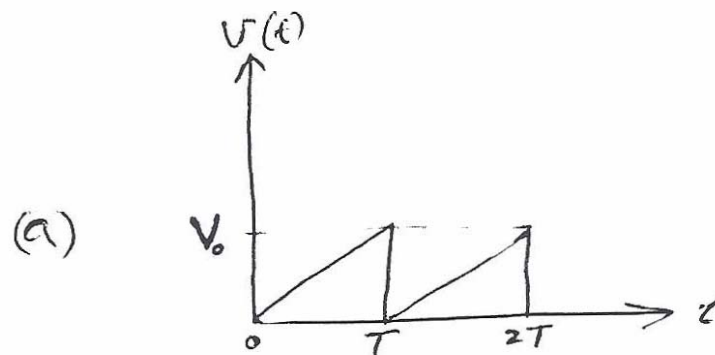
Voltmeter A with a sensitivity of  $1000\ \Omega/\text{V}$  and Voltmeter B with a sensitivity of  $20,000\ \Omega/\text{V}$ .



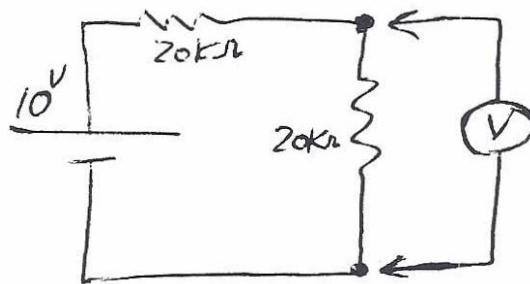
Both meters have 0-50V range. Calculate (a) the reading of each voltmeter and (b) the error in each reading expressed as a percentage of true V.

4-8

Two generators with  $500\ \Omega$  internal resistance have a SAWTOOTH and Tropicoidal output voltages as shown. The r.m.s. values of these outputs to be measured by a moving coil instrument whose internal resistance is  $10\text{ k}\Omega$ . The instrument has a full wave rectifier and is calibrated for sinusoidal wave forms. Calculate the errors due to the waveform and also the loading errors.



- 4-9** A voltmeter has a resistance of  $20 \text{ K}\Omega/\text{V}$ ; is used to measure the voltage on the shown circuit on a  $10\text{V}$  range. Find the percentage loading error.



- 4-10** A voltage source of an internal resistance  $R_0 = 0.4 \text{ K}\Omega$  is measured 5 times with a moving coil meter. The following readings were obtained

3.21 , 3.22 , 3.20 , 3.23 , 3.24 V

The same source was also measured by a recently calibrated digital voltmeter whose input resistance is  $10 \text{ M}\Omega$ . The following 4 readings were obtained

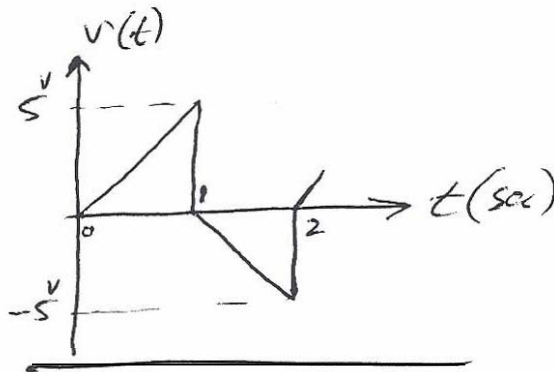
3.412 , 3.413 , 3.411 , 3.412 V

- Estimate the percentage loading error of the digital voltmeter.
- Find the precision and accuracy of both sets of readings.
- Does any of the readings need recalibration? How ?
- If the bias in the moving coil meter is due solely to its loading error, estimate its input resistance.

- 4-11** A moving coil movement has 100 turns,  $5 \text{ cm}^2$  coil area, and air-gap magnetic flux density of  $0.1 \text{ t (Wm}^{-2}\text{)}$ . The control spring exerts a torque of  $5 \times 10^{-6} \text{ Nm}$  at the full-scale deflection of  $90^\circ$ . The potential difference across the coil terminals at full-scale deflection is  $100 \text{ mV}$ . Using the above movement, design a multi-range DC ammeter with ranges  $[0, 50] \text{ mA}$  and  $[0, 1] \text{ A}$ , and a multi-range DC voltmeter with ranges  $[0, 10] \text{ V}$  and  $[0, 200] \text{ V}$ .

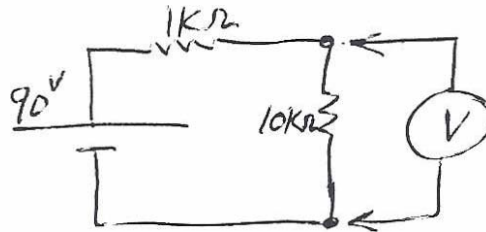
- 4-12** An average reading full-wave rectifier moving coil AC voltmeter is calibrated to read correctly the RMS value of applied sinusoidal voltages. The periodic waveform  $V(t)$  shown is applied to the meter

Calculate  $V_{RMS}$  for this waveform, Vindicated and the waveform error in it



**4-13** A D'Arsonval movement gives full scale deflection of 1. mA when a voltage of 50 mV is applied across its terminals.

Calculate the resistance that should be added in series with this movement to convert it into a 0-100 V voltmeter. The above 0-100 V voltmeter is used to measure the voltage across the 10 KO resistor in the shown circuit. Determine the percentage loading error.



**4-14** For a D'Arsonval movement, the reading is  $\theta = 45^\circ \pm 0.5^\circ$ . If the number of turns is  $N = 100 \pm 0.3$ , the coil area is  $A = 6 \text{ cm}^2 \pm 1\%$ , the magnetic flux density is  $B = 0.2 \text{ t} \pm 1\%$ , and the total spring constant is  $K = 5 \times 10^{-8} \text{ Nm/o} \pm 2\%$ ,

**Find**

- the nominal value and percentage uncertainty of the current I.
- the nominal value of the movement sensitivity.

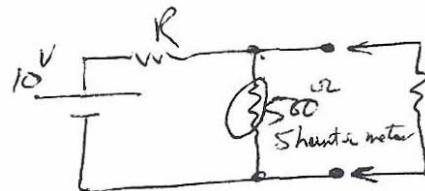
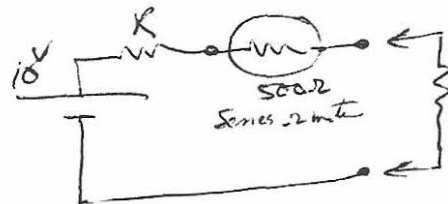


**4-15** A D'Arsonval movement has a full-scale deflection current,  $I_{FSD} = 50$  mA and a coil resistance  $R_m = 50 \Omega$ . Use this movement to design

- an ammeter with range [0, 2] A
- a voltmeter with range 10, 20 V.

**4-16** What would you mark as readings in the 0-scale for the shown 0-meter circuits when the readings in the current scale is :

- a)  $I_{FSD} = 100 \mu\text{Amp}$   
 b)  $50 \mu\text{A}$   
 c)  $30 \mu\text{Amp}$   
 d)  $10 \mu\text{Amp}$



**4-17** Find the steady state deflection of a moving coil galvanometer in degrees and in mm on a scale placed 1 m away when a current of 1 mA is passed through its coil. Also calculate the resistance required for critical damping if the galvanometer has a relative damping of 0.1 on open circuit. The data of the galvanometer is :

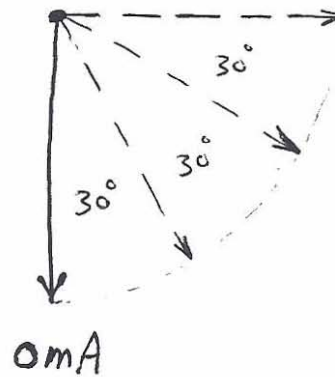
Dimensions of coil =  $25 \times 20$  mm, number of turns = 100, flux density =  $0.1 \text{ Wb/m}^2$ , inertia constant =  $0.1 \times 10^{-6} \text{ kg-m}^2$  and control constant =  $25 \times 10^{-6} \text{ Nm/rad}$ .

**4-18** The fig. shows a circuit for testing of a galvanometer where  $E = 1.5 \text{ V}$ ,  $R_1 = 1.0 \Omega$ ,  $R_2 = 2500 \Omega$ ,  $R_3$  is variable.

When  $R_3 = 450 \Omega$ , the galvanometer deflection is 150 mm and with  $R_3 = 950 \Omega$ , the deflection is reduced to 75 mm. Calculate (a) the resistance of galvanometer, (b) current sensitivity of galvanometer.



4-19 An MCV has a gravitational restoring force with zero position of needle as shown in the figure. The movement parameters are:



$$N = 100 \text{ Turn,}$$

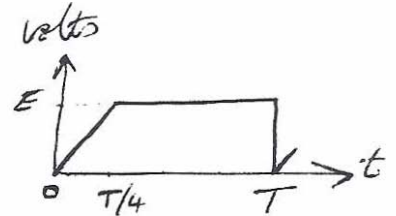
$$B = 0.3 \text{ Tesla,}$$

$$A = 4 \text{ cm}^2,$$

$$m = 10 \text{ grams } \neq$$

$C_g = 2 \text{ cm}$  below pivot. Find the readings at  $30^\circ$  spaced positions off zero. What is  $i_{FSD}$ ? Why?

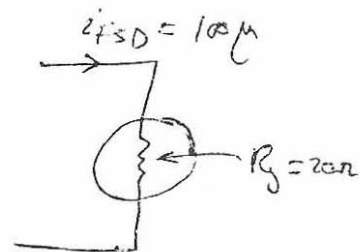
4-20 An average half-wave rectifier AC MCV is calibrated to read correctly rms of sine signals.



Find %e in reading the ac of the signal shown above.

4-21 A MCV has  $R_g = 20 \Omega$ ,  $i_{FSD} = 100 \mu A$  was required

to measure  $10 A$  &  $300 V$  ranges

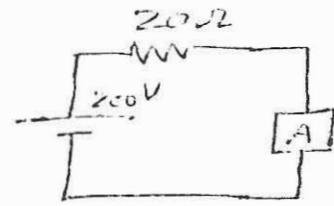


a) Design to match these purposes.

b) What is the uncertainty of scale at maximum deflection if:  $\Delta R_g = 2\%$ ,  $\Delta i_{FSD} = 1 \mu A$  & uncertainty in all used resistors is  $10\%$ .

**4-22** Design a 10 Amp-range ammeter given a movement with  $R_m = 100 \Omega$  &  $I_{FSD} = 100 \mu\text{Amp}$ .

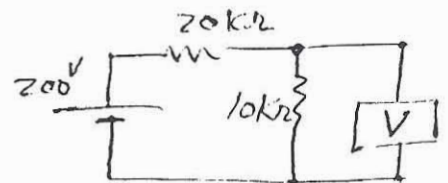
- a) What is the reading using your ammeter if it was to measure the current through the circuit given aside?



- b) What is the bias % error? What causes this error?
- c) Which range of 20A, 200A or 2000A ranges of 3 1/2 digit DMM with  $50 \text{ mV/Amp}_{FSD}$  would you rather select? Why?

**4-23** Design a 100-DCV range voltmeter given a movement with  $R_m = 100 \Omega$  &  $I_{FSD} = 100 \mu\text{Amp}$ .

- a) What is the reading using your voltmeter if it was to measure the voltage on the circuit given aside?!



- b) What is the bias % error?!
- c) Which range of 60V, 70V, 80V ranges of 3 digit DMM with  $1 \text{ K}\Omega/\text{V}_{FSD}$  would you select, why?

# Problem Set Five, Moving Iron Meter.

**S-1** A moving iron voltmeter is used to measure the rms of the following signal:

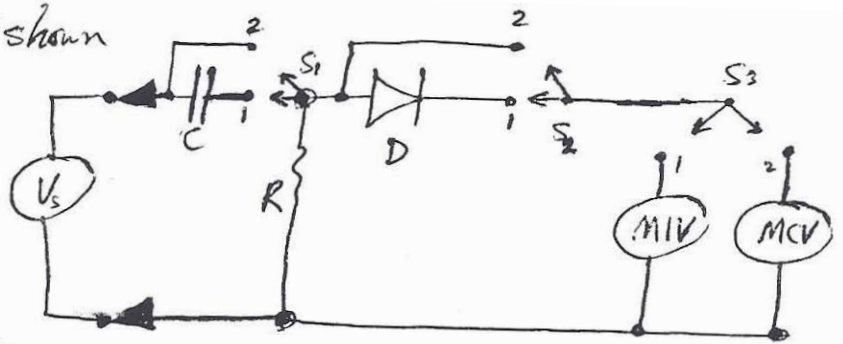
$$V(t) = 5 + 12 \sin \omega t + 12 \cos 2\omega t$$

What do you expect to give as reading.

Suppose now that the moving coil voltmeter were used what is the reading you expect to see? Which meter you prefer to use & why?

**S-2** For the circuit shown

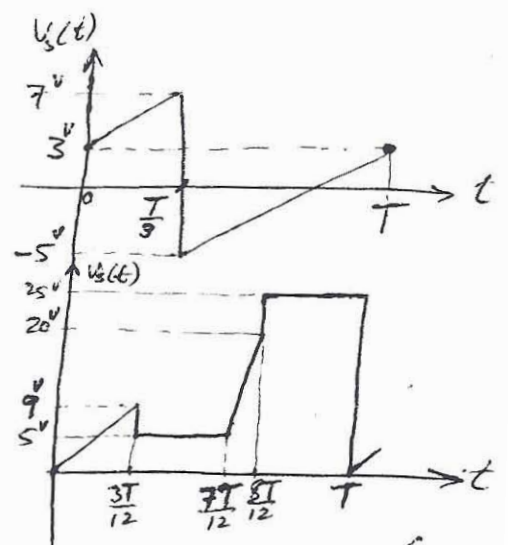
in the fig, assuming ideal filter and diode, the switches  $S_1, S_2$  &  $S_3$



enable 4 readings for both MIV & MCV. Which reading is representative for the rms of  $V_s$ , if  $V_s$  is given by:

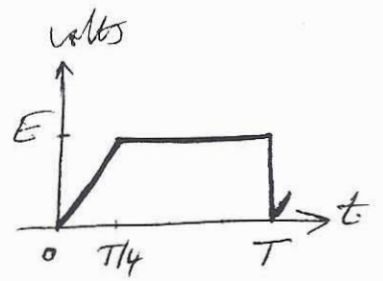
- $V_s(t) = 8$  volts
- $V_s(t) = 12 \sin \omega t$
- $V_s(t) = 8 + 12 \sin \omega t$
- $V_s(t)$  the waveform shown in Figs.

Calculate the errors and give your comments to every reading.

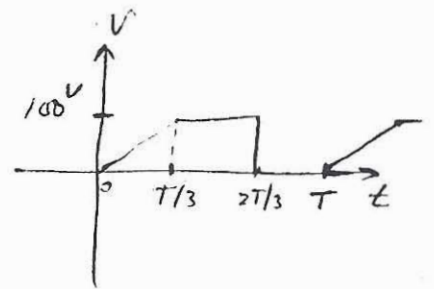




**5-3** A source with  $10\Omega$  internal resistance has the pattern shown in the figure. An MIV with  $R_g = 1K\Omega$  was used to measure its rms value. Find the reading & %e. Comment.

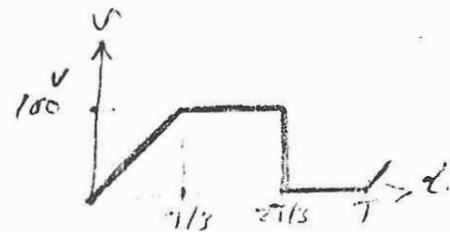


**5-4** Two voltmeters with very high input resistance, MCV & MIV, are used to measure the rms of the voltage signal given here.



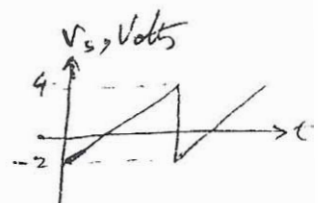
- Indicate the reading of both,
- which one has error, how much is it?
- what is the cause of it.

**5-5** Two voltmeters with very high input resistances, MCV & MIV, were used to measure the dc of this signal.



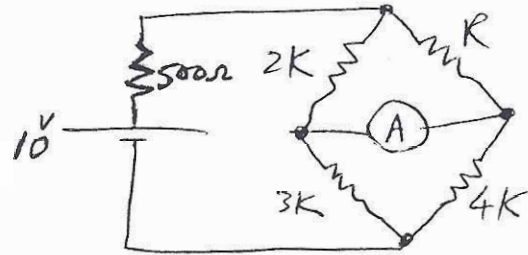
- Indicate the reading of both.
- How would you get your measurement?
- What is the error in it & f. who causes it

**5-6** Two voltmeters MCV & MIV were used to measure the rms of the signal shown find readings and errors



## Problem Set Six, Bridges.

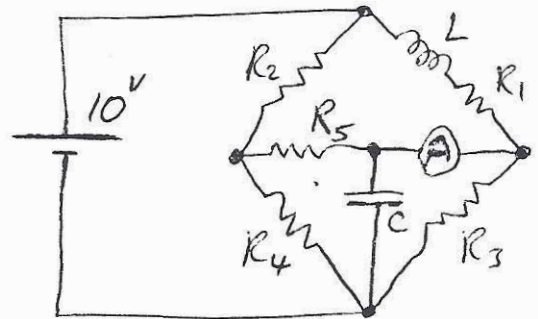
**6-1** What is the unknown resistance,  $R$  in the shown Wheatstone Bridge.



**6-2** If  $R$  is varied by  $+2\%$ , what would be the Ammeter reading. Take movement resistance of  $100\Omega$ .

**6-3** In the Anderson Bridge shown

the ac source was made a dc one. What are the conditions of balance?!



If  $R_1 = R_3 = 10\text{K}\Omega$  &

$R_2 = R_4 = R_5 = 12\text{K}\Omega$  &

$C = 10\mu\text{F}$  &  $L = 50\text{mH}$ , then find the Ammeter reading if  $R_m = 200\Omega$ .

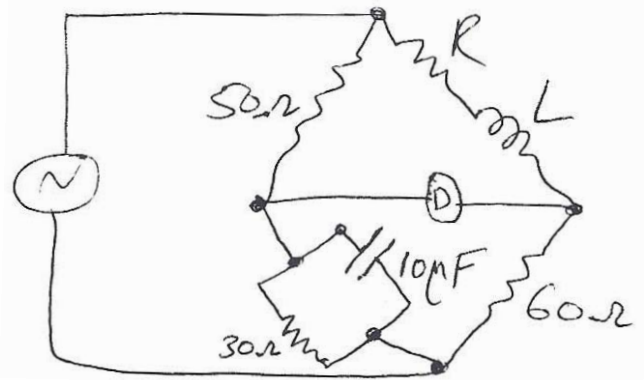
Repeat the calculations for  $+10\%$  change in  $R_2$ .

Find the voltages across both  $L$  &  $C$  in both cases.

**6-4** The above Anderson Bridge was fed from an ac source. What are the values of  $R_2$  &  $C$  that are required for equilibrium?



6-5 Find the value  
of  $R$  &  $L$   
for the shown  
Maxwell's Bridge.

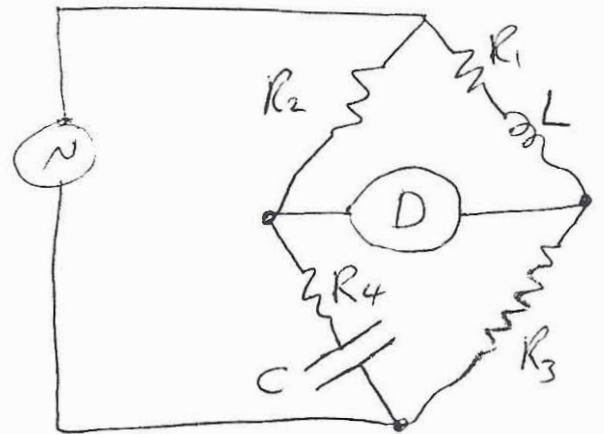


6-6 Derive the balance  
conditions for Hay's  
Bridge. Does it change  
with frequency?

Find  $L$  &  $C$  if

$$R_1 = 50\Omega, R_2 = 120\Omega$$

$$R_3 = 15\Omega, R_4 = 80\Omega. \text{ Use mains frequency of } 60\text{Hz.}$$

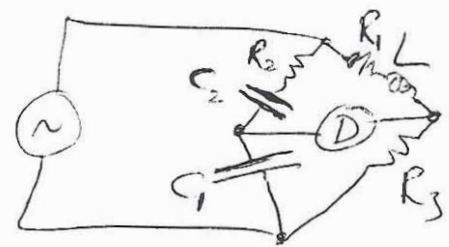


6-7 Derive the balance  
conditions for Owen's Bridge.  
Is it affected by frequency?

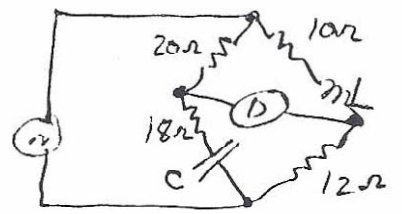
Find  $C_2$  &  $R_2$  if

$$R_1 = 1\text{K}\Omega, L = 50\text{mh}, R_3 = 500\Omega \text{ \& } C_1 = 40\mu\text{F}$$

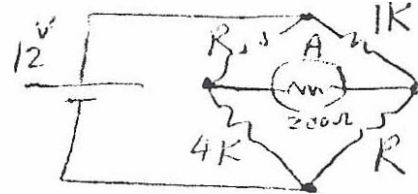
with mains frequency.



**6-8** What are values of  $L$  &  $C$  for the given Hay's bridge at balance? assume main frequency of  $60 \text{ Hz}$ .



**6-9** Two identical thermometers with coefficient of  $.01/^\circ\text{C}$  keeps the shown bridge balanced at an ambient of  $25^\circ\text{C}$ . What would be the temperature to cause  $20 \mu\text{A}$  deflection?



## Problem Set Seven, Wattmeter.

**7-1** A dynamometer wattmeter, having a current coil resistance of  $0.5 \Omega$  and a voltage coil resistance of  $12.5 \text{ k}\Omega$  was used to measure the power in  $250 \text{ DCV}$  load at (a)  $4 \text{ Amps}$  & (b)  $12 \text{ Amps}$ .

Which connection would you use for each case, why? Also find % errors.

---

**7-2** If the reactance of the current coil of the above wattmeter was  $1\%$  of its resistance, repeat the answer for  $250 \text{ ACV}$  load at unity pf for (a)  $4 \text{ Amps}$  & (b)  $12 \text{ Amps}$

---

**7-3** Repeat the above problem for pf of (a)  $0.8$ , (b)  $0.5$  & (c)  $0.1$

---

**7-4** Repeat exercises 2 & 3 if the reactance of the voltage coils is  $0.1\%$  of its resistance.

---

7-5 How can we measure the power delivered to a 3- $\phi$  load connected (a) in star & (b) in  $\Delta$  using: (I) one wattmeter (II) two wattmeters (III) three wattmeters. State also the conditions of measurement.

---

7-6 Can you measure the reactive power in 3- $\phi$  load using one wattmeter? If yes state how and conditions of measurement:

- (a) If load is Y.
  - (b) If load is  $\Delta$ .
-

## Problem Set Eight, Power-Factor Meter

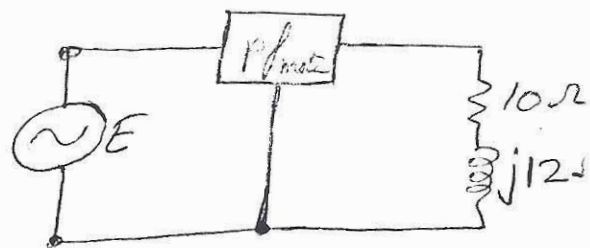
**8-1** A pf-meter has one voltage coil impedance of  $(20,000 + j \cdot 1) \Omega$ , find the voltage meter-coil impedance.

If the applied input frequency was 10% higher than the designed one, how would you compensate the first coil.

---

**8-2** The above pf meter was put across the load in the given circuit,

Find the reading and the percentage error in it for



both possible connections if the current coil impedance was  $(0.2 + j0.5) \Omega$ . Which connection do you prefer?

**8-3** If, by mistake, the voltage & current coils were swapped over, find the two possible readings. Comment.

**8-4** If, by ignorance, the voltage & current coils were shorted together (a) across the load & (b) in series with it, find the readings and give your comments & explanation

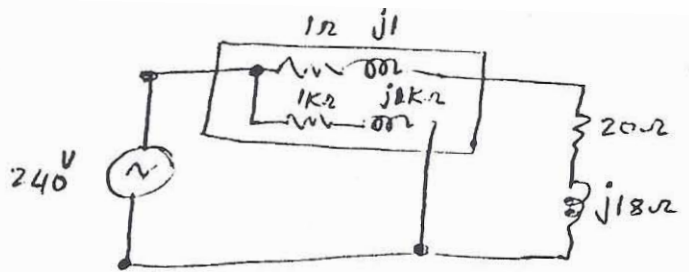


8-5 Find the reading of

% error for the 1- $\phi$

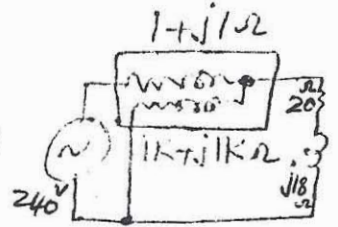
(a) wattmeter shown.

(b) pf-meter shown



8-6 Find the reading of % error for the shown 1- $\phi$

(a) wattmeter + (b) pf meter



8-7 Suppose that the impedance of voltage coil-pairs of a pf meter were not equal and were  $Z_R \angle \theta_R$  for the resistive dominated coil and  $Z_L \angle \theta_L$  for the inductive dominated coil. Prove that the angle of deflection,  $\psi$ , is given by:

$$\psi = \tan^{-1} \left[ \frac{Z_R \cos(\phi - \theta_L)}{Z_L \cos(\phi - \theta_R)} \right]$$

where  $\phi$  is the actual power factor angle.

$$\text{If } Z_R = R \angle 0^\circ \quad \phi$$

$$Z_L = R \angle 90^\circ$$

Show that the above formula becomes:

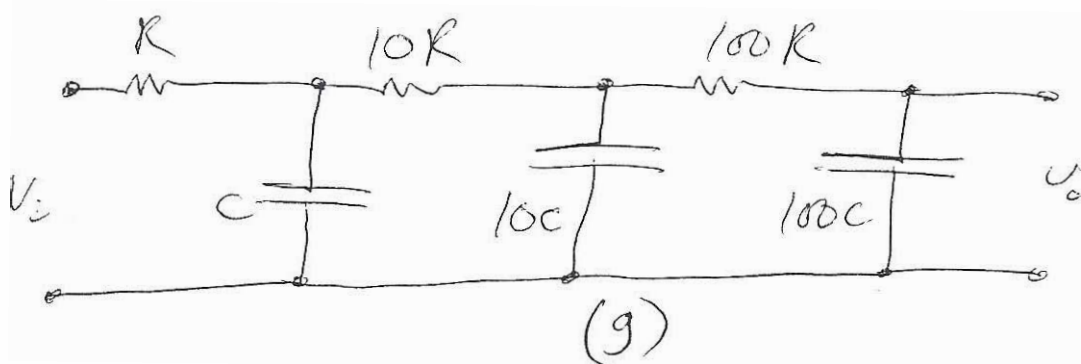
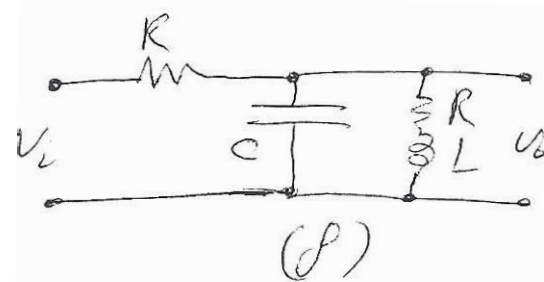
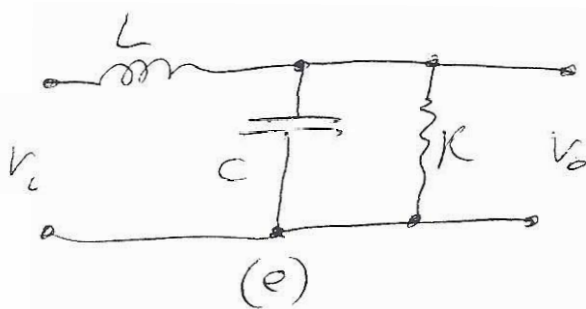
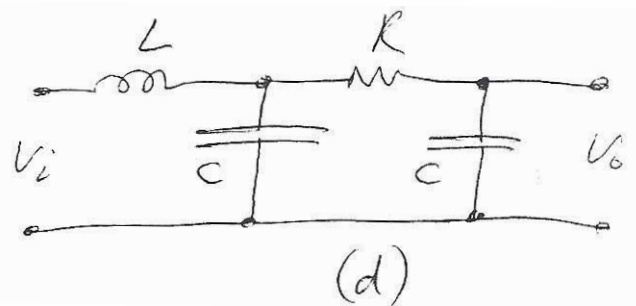
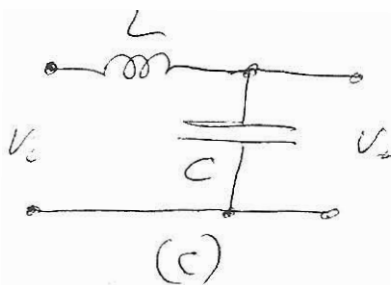
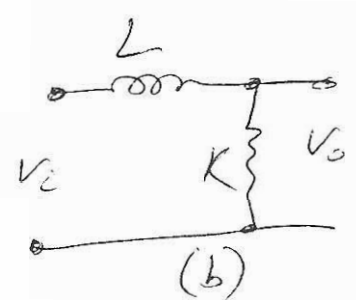
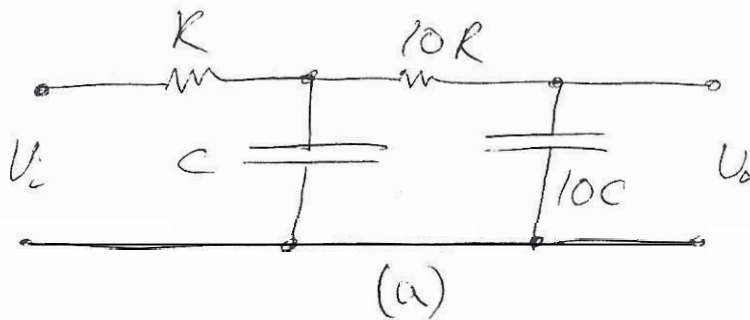
$$\psi = \phi$$

and hence the pf meter measures correctly.

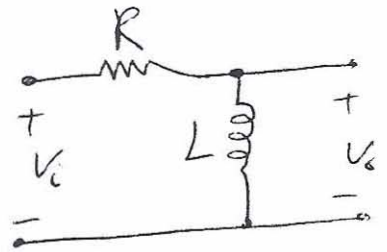


# Problem Set Nine, Filters.

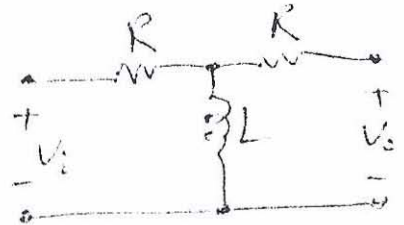
**9-1** Find the Bandwidth of the following networks, and their dc gains



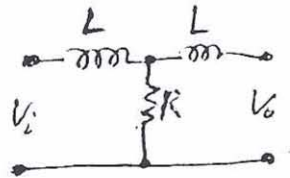
9-2] What type of filter is the one shown in the figure. Find  $f_L$  &  $f_H$  & gain at  $f = 2f_L$ .



9-3] Classify the filter and find its  $f_L$  &  $f_H$  given  $R = 10 \Omega$ ,  $L = 10 \text{ mH}$



9-4] Find the bandwidth of the shown network, given  $L = 10 \text{ mH}$  &  $R = 10 \Omega$ .



$$\boxed{1-1} \quad \therefore C = \frac{\epsilon A}{d} = \frac{\epsilon_r \epsilon_0 A}{d}$$

$$\therefore A = 5 \text{ cm}^2 \pm 0.5\%$$

$$\therefore \frac{dA}{A} = 0.005$$

$$\# \quad C = 950 \text{ pF}$$

$$\# \quad \therefore d = \frac{\epsilon_r \epsilon_0 A}{C} = \frac{8.854 \times 10^{-12} \times 81 \times 5 \times (10^2)^2}{950 \times 10^{-12}}$$

$$= 3.775 \times 10^{-4} \text{ m} = 0.38 \text{ mm, the separation.}$$

$$\# \quad \frac{\partial C}{\partial d} = \frac{-\epsilon A}{d^2} = \frac{-\epsilon_r \epsilon_0 A}{d \cdot d} = \frac{-C}{d} =$$

$$= \frac{-950 \text{ pF}}{0.38 \text{ mm}} = -2.52 \text{ nF/mm, the sensitivity.}$$

$$\# \quad \left(\frac{dc}{c}\right)^2 = \left(\frac{dA}{A}\right)^2 + \left(\frac{dd}{d}\right)^2 \quad \therefore \frac{dd}{d} = 3\% = 0.03$$

$$\therefore \frac{dc}{c} = \sqrt{(0.005)^2 + (0.03)^2} = 0.0304 = 3.04\%$$

$\therefore$  The maximum uncertainty in  $c$  is  $3.04\%$

$$\boxed{1-2} \text{ a) } T = \frac{T_0}{1 + (T_0/\beta) \ln(R/R_0)} = \frac{300}{1 + (300/3420) \ln(2/1)} = 282.8 \text{ } ^\circ\text{K}$$

$$\text{b) } S = \frac{\partial R}{\partial T} = R \beta_0 \left(-\frac{1}{T^2}\right) = -\frac{\beta_0 R}{T^2} = -\frac{3420 * 2\text{K}}{(282.8)^2} = -85.5 \text{ } \Omega/\text{K}$$

$$\text{c) } \therefore R = R_0 \exp\left(\beta_0 \left(\frac{1}{T} - \frac{1}{T_0}\right)\right)$$

$$\begin{aligned} \therefore dR &= R_0 dR_0 + R_T \cdot dT = (R/R_0) dR_0 + R \cdot \beta_0 \left(-\frac{1}{T^2}\right) dT = \\ &= R \cdot \left[ \frac{dR_0}{R_0} - \frac{\beta_0}{T} \cdot \frac{dT}{T} \right] \end{aligned}$$

$$\therefore \frac{dR}{R} = \frac{dR_0}{R_0} - \frac{\beta_0}{T} \cdot \frac{dT}{T} \Rightarrow \frac{\beta_0}{T} \cdot \frac{dT}{T} = \frac{dR_0}{R_0} - \frac{dR}{R}$$

$$\therefore \frac{dT}{T} = \frac{T}{\beta_0} \cdot \left( \frac{dR_0}{R_0} - \frac{dR}{R} \right) \quad (\text{vector equation.})$$

$$\therefore \left(\frac{dT}{T}\right)^2 = \left(\frac{T}{\beta_0}\right)^2 \left[ \left(\frac{dR_0}{R_0}\right)^2 + \left(\frac{dR}{R}\right)^2 \right] \quad \therefore \text{OK f max } dT = 0.167\% = 0.473$$

$$\boxed{1-3} \text{ a) } MV_{mcm} = 6.72 \text{ Volts}$$

$$MV_{DVM} = 7.072 \text{ Volts} = TV \quad \text{since Loading is small \& it is recently calibrated.}$$

$$\therefore \text{Prec.}_{mcm} = |6.72 - 6.69| = 0.03 \text{ Volts.}$$

$$\& \text{ Acc.}_{mcm} = |7.072 - 6.69| = 0.382 \text{ Volts.}$$

$$\text{Prec.}_{DVM} = \text{Acc.}_{DVM} = |7.072 - 7.069| = 0.003 \text{ Volts}$$

b) The average of MCV should be increased by the bias =  $7.072 - 6.72 = 0.352$  Volts, to be true, and so is every reading of mcv.

$$\boxed{1-4} \text{ a) } W = \frac{E X^4 Y}{8 D^3 N}$$

$$\therefore W = \frac{80 \times 10^9 \times (2.5 \times 10^{-3})^4 (10 \times 10^{-2})}{8 (2.5 \times 10^{-2})^3 \times 50} = 50 \text{ N}$$

$$\text{b) } \frac{dW}{W} = \frac{dE}{E} + 4 \frac{dX}{X} + \frac{dY}{Y} - 3 \frac{dD}{D} - \frac{dN}{N} \quad (\text{Vectorially})$$

$$\therefore \left| \frac{dW}{W} \right| = \sqrt{\left(\frac{dE}{E}\right)^2 + 16 \left(\frac{dX}{X}\right)^2 + \left(\frac{dY}{Y}\right)^2 + 9 \left(\frac{dD}{D}\right)^2 + \left(\frac{dN}{N}\right)^2} =$$

$$= \sqrt{16 + 16 + .04 + 36 + 1} \% = \sqrt{69.04} \% = 8.31 \%$$

$$\therefore dW = 4.15 \text{ N}$$

∴ order of contribution as seen from the above figures is  $D, E \& X, N, Y$

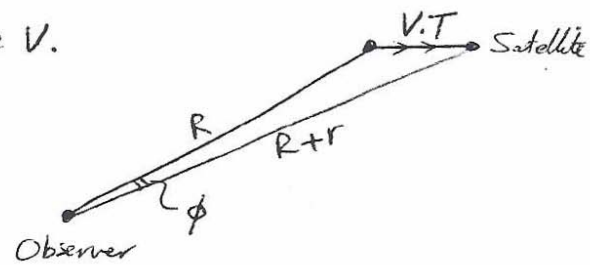
$$\text{c) } S = \frac{\partial Y}{\partial W} = \frac{8 D^3 N}{E X^4} = \frac{8 (.025)^3 (50)}{80 \times 10^9 \times (.0025)^4} = 0.002 \text{ m/N}$$

The sensitivity is independent upon  $W$  ∴ hence the plot is constant line

$\boxed{1-5}$  Assume the speed of satellite to be  $V$ .

∴ time between observations to be  $T$

$$\begin{aligned} \therefore (VT)^2 &= R^2 + (R+r)^2 - 2R(R+r) \cos \phi \\ &= 2R^2 + 2rR + r^2 - 2R^2 \cos \phi - 2Rr \cos \phi \\ &= r^2 + 2R(R+r)(1 - \cos \phi) \end{aligned}$$



$$\therefore V^2 = \frac{r^2 + 2R(R+r)(1 - \cos \phi)}{T^2}$$

$$\therefore V = 51.56 \text{ Km/sec}$$

$$\text{∴ } 2VdV = \frac{1}{T^2} \cdot \left[ (2r + 2R(1 - \cos \phi)) dr + (1 - \cos \phi)(4R + 2r) dR + 2R(R+r) \sin \phi d\phi \right]$$

$$\therefore |dV| = \frac{1}{VT^2} \sqrt{\left\{ (r + R(1 - \cos \phi)) dr \right\}^2 + \left\{ (1 - \cos \phi)(2R + r) dR \right\}^2 + \left\{ R(R+r) \sin \phi d\phi \right\}^2}$$

$$= 0.1167 \text{ Km/sec} = 116.7 \text{ m/sec}$$

∴ The satellite was moving at  $51.56 \text{ Km/sec} \pm 116.7 \text{ m/sec}$



$$\boxed{1-6} \quad \text{True Val.} = 340 \text{ } \mu\text{m/m}$$

$$\# \quad \text{Av.} = \frac{\sum \text{all readings}}{12} = \frac{4134}{12} = 344.5 \text{ } \mu\text{m/m}$$

$$\# \quad \therefore \text{Bias} = \text{TV} - \text{AV} = 340 - 344.5 = -4.5 \text{ } \mu\text{m/m}$$

$$\# \quad \text{Precision} = |\text{AV} - \text{Max. Deviated Reading}| =$$

$$= |344.5 - 351.0| = 6.5 \text{ } \mu\text{m/m}$$

$$\# \quad \text{Accuracy} = |\text{TV} - \text{Max. Deviated Reading}| =$$

$$= |340.0 - 351.0| = 11.0 \text{ } \mu\text{m/m}$$

$$\boxed{1-7} \quad z = P \cdot y_1^{i_1} \cdot y_2^{i_2} + S$$

$$\therefore z - S = P y_1^{i_1} \cdot y_2^{i_2}$$

$$\therefore \frac{d(z - S)}{(z - S)} = \frac{dP}{P} + \frac{i_1 dy_1}{y_1} + \frac{i_2 dy_2}{y_2} + (\ln y_1) \frac{di_1}{i_1} + (\ln y_2) \frac{di_2}{i_2}$$

$$\frac{dz - dS}{z - S} = \frac{w_0}{1} + 1 \cdot \frac{w_{10}}{y_{10}} + 1 \cdot \frac{w_{20}}{y_{20}} + (\ln y_{10}) \frac{w_{11}}{1} + (\ln y_{20}) \frac{w_{21}}{1}$$

$$\therefore \frac{dz - w_3}{(1 \cdot y_{10}^{i_1} \cdot y_{20}^{i_2} + 0) - 0} = w_0 + \frac{w_{10}}{y_{10}} + \frac{w_{20}}{y_{20}} + (\ln y_{10}) w_{11} + (\ln y_{20}) w_{21}$$

$$\therefore dz = w_3 + y_{10}^{i_1} \cdot y_{20}^{i_2} \left[ w_0 + \left(\frac{1}{y_{10}}\right) w_{10} + \left(\frac{1}{y_{20}}\right) w_{20} + (\ln y_{10}) w_{11} + (\ln y_{20}) w_{21} \right]$$

$$\therefore |dz| = \sqrt{w_3^2 + y_{10}^2 \cdot y_{20}^2 \cdot \left( w_0^2 + \frac{w_{10}^2}{y_{10}^2} + \frac{w_{20}^2}{y_{20}^2} + w_{11}^2 \ln^2 y_{10} + w_{21}^2 \ln^2 y_{20} \right)}$$



$$\boxed{1-8} \quad R = R_0 (1 + \alpha (T - T_0))$$

$$\# \therefore T = \left[ \left( \frac{R}{R_0} - 1 \right) / \alpha \right] + T_0$$

$$= T_0 + \frac{R - R_0}{\alpha R_0} \quad \text{assume } T_0 = 25^\circ\text{C} \\ = (273 + 25)^\circ\text{K} \\ = 298^\circ\text{K}$$

$$\# \quad T = 298 + \frac{30 \times 10^3 - 20 \times 10^3}{.00392 \times 20 \times 10^3} = 425.6^\circ\text{K} \\ = 152.6^\circ\text{C}$$

$$\# \quad \frac{\partial R}{\partial T} = R_0 \alpha = 20 \times 10^3 \times .00392 = 78.4 \Omega / ^\circ\text{K}$$

$$\# \quad dR = [1 + \alpha (T - T_0)] dR_0 + R_0 \alpha dT = \frac{R dR_0}{R_0} + \alpha R_0 dT$$

$$\therefore dT = \frac{dR}{\alpha R_0} - \frac{R dR_0}{R_0^2 \alpha} = \frac{R}{\alpha R_0} \left[ \frac{dR}{R} - \frac{dR_0}{R_0} \right]$$

$$\therefore |dT| = \sqrt{(dR/R)^2 + (dR_0/R_0)^2} \cdot \frac{R}{\alpha R_0}$$

$$= \sqrt{(.001)^2 + (.001)^2} \cdot \frac{30}{20 \times .00392} = .541^\circ\text{K}$$

$\therefore$  Uncertainty in  $T$  assuming  $T_0$  to be accurate is  $0.541^\circ\text{K}$

i.e.  $0.127\%$

$$\# \quad dT(T_0, R_0, R) = (1) dT_0 + \left( \frac{-R}{R_0^2 \alpha} \right) dR_0 + \left( \frac{1}{\alpha R_0} \right) dR$$

$$\therefore |dT|^2 = |dT_0|^2 + \left( \frac{R}{\alpha R_0} \right)^2 \left( \left| \frac{dR_0}{R_0} \right|^2 + \left| \frac{dR}{R} \right|^2 \right)$$

1-9

Let  $R_1 = 20\Omega \pm 5\%$


$\therefore dR_1/R_1 = .05$

$\neq dR_1 = 1\Omega$

$\neq$  let  $R_2 = 10\Omega \pm 10\%$

$\therefore dR_2/R_2 = .1$

$\neq dR_2 = 1\Omega$

# Connection A:  $\therefore R_A = R_{a1} + R_{a2} + R_{a3}$  

$\therefore dR_A = dR_{a1} + dR_{a2} + dR_{a3}$

$\therefore |dR_A|^2 = (dR_{a1})^2 + (dR_{a2})^2 + (dR_{a3})^2 = 3(dR_2)^2 = 3 \times 1^2$

$\therefore |dR_A| = \sqrt{3} \Omega$

# Connection B:  $\therefore R_B = R_{b1} + R_{b2}$

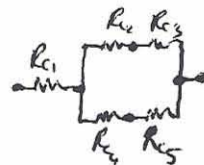


$\therefore dR_B = dR_{b1} + dR_{b2}$

$\therefore |dR_B|^2 = |dR_{b1}|^2 + |dR_{b2}|^2 = |dR_1|^2 + |dR_2|^2 = 1^2 + 1^2 = 2$

$\therefore |dR_B| = \sqrt{2} \Omega$

# Connection C:  $\therefore R_C = R_{c1} + \frac{(R_{c2} + R_{c3})(R_{c4} + R_{c5})}{R_{c2} + R_{c3} + R_{c4} + R_{c5}}$



$\therefore dR_C = dR_{c1} + \left[ \frac{(R_{c4} + R_{c5}) - (R_{c2} + R_{c3})(R_{c4} + R_{c5})}{(R_{c2} + R_{c3} + R_{c4} + R_{c5})^2} \right] dR_{c2} + \text{similar terms for } dR_{c3}, dR_{c4}, dR_{c5}$

$\therefore |dR_C|^2 = |dR_{c1}|^2 + \left(\frac{R_{c4} + R_{c5}}{\Sigma}\right)^4 \left[ |dR_{c2}|^2 + |dR_{c3}|^2 \right] + \left(\frac{R_{c2} + R_{c3}}{\Sigma}\right)^4 \left[ |dR_{c4}|^2 + |dR_{c5}|^2 \right]$

$= |dR_1|^2 + \left(\frac{R_2 + R_2}{4R_2}\right)^4 \left[ |dR_2|^2 + |dR_2|^2 \right] + \left(\frac{R_2 + R_2}{4R_2}\right)^4 \left[ |dR_1|^2 + |dR_2|^2 \right]$

$= |dR_1|^2 + \left(\frac{1}{2}\right)^4 \cdot 4 |dR_2|^2 = |dR_1|^2 + \frac{1}{4} |dR_2|^2 = 1^2 + \frac{1}{4} = 1.25 = \frac{5}{4}$

$\therefore |dR_C| = \frac{\sqrt{5}}{2} = \sqrt{1.25} = 1.12 \Omega$

$\therefore \sqrt{3}, \sqrt{2} \neq \sqrt{1.25}$ , The best is connection C with  $|dR_C| = 1.12\Omega$

$$\boxed{1-10} \quad T_{av} = (T_1 + T_2 + \dots + T_N) / N$$

$$\therefore dT_{av} = (dT_1 + dT_2 + dT_3 + \dots + dT_N) / N$$

$$\begin{aligned} \therefore |dT_{av}|^2 &= \left[ |dT_1|^2 + |dT_2|^2 + \dots + |dT_N|^2 \right] / N^2 \\ &= N |dT|^2 / N^2 = |dT|^2 / N \end{aligned}$$

$$\therefore |dT_{av}| = \frac{|dT|}{\sqrt{N}}$$

$\therefore$  Uncertainty of measurement can be reduced by a factor of  $\frac{1}{\sqrt{N}}$  if the average of  $N$  readings was taken.

---

2-2 a)  $V_1 = 40 \sin(628t + \pi/5) \text{ V}$

$\therefore Y_1 = 2 \sin(2\pi \times 100t + \pi/5) \text{ Cm}$

$V_2 = 60 \sin 628t \text{ V}$

$\therefore Y_2 = 3 \sin(2\pi \times 100t) \text{ Cm}$

frequency of both signals = 100 Hz

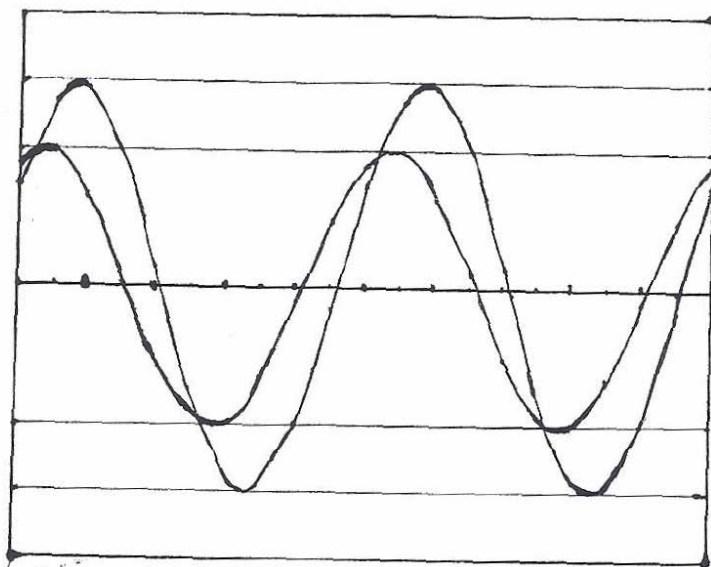
$\therefore$  Period = 10 msec = 5 Cm

$\therefore$  Trig. level is 30 V of ch2 with ~~the~~ slope

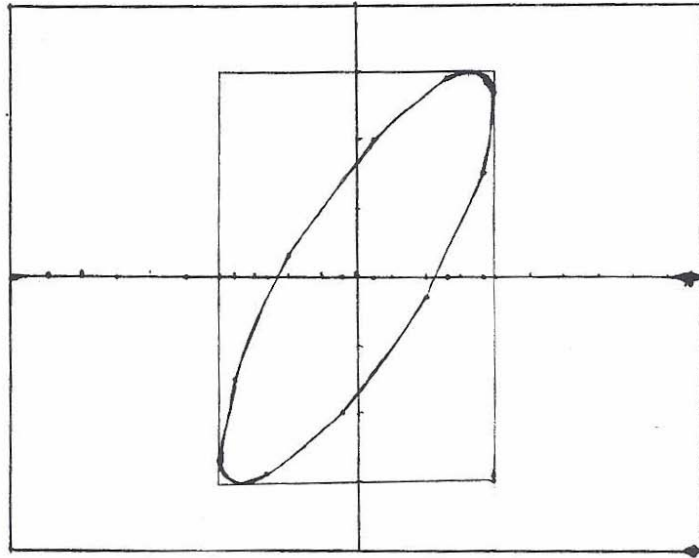
$\therefore$   $Y_2$  will start at left of screen at 1.5 Cm heading up.

$\therefore$  The trace will be as follows, for balanced display:


$\frac{\omega t}{2\pi}$	$Y_1(\text{Cm})$	$Y_2(\text{Cm})$
5/60	1.8	1.5
11/60	2.0	2.7
17/60	1.3	2.9
23/60	0.2	2.0
29/60	-1.0	0.3
35/60	-1.8	-1.5
41/60	-2.0	-2.7
47/60	-1.3	-2.9
53/60	-0.2	-2.0
59/60	+1.0	-0.3
65/60	1.8	1.5




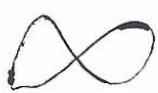
b) Now  $Y_1$  stands for  $X$  &  $Y_2$  for  $Y$ , hence with  $X, Y$  mode the pattern will look like this:



c) Lissajous figures:

I)   $\therefore \frac{1}{2} f_h = \frac{1}{2} f_v \quad \therefore f_v = f_h = 1 \text{ KHz}$

II)   $\therefore 1 \cdot f_h = 1 \frac{1}{2} f_v \quad \therefore f_v = \frac{2}{3} f_h = \frac{2}{3} \text{ KHz}$

III)   $\therefore 2 f_h = 1 \cdot f_v \quad \therefore f_v = 2 f_h = 2 \text{ KHz}$

IV)  $\therefore 3 f_h = 5 f_v \quad \therefore f_v = \frac{3}{5} f_h = 600 \text{ Hz}$



**2-6** From trace  $\therefore$  Period =  $7.2 \text{ cm} = 7.2 \times 10 \text{ ms} = 72 \text{ ms}$   
 $\therefore f = \frac{1}{T} \approx 14 \text{ Hz}$   $\therefore \omega = 2\pi f = 2\pi \times 14 \approx 88 \text{ rad/sec}$   
 $\therefore Y_1 = 3 \text{ cm peak}$  &  $Y_2 = 3.8 \text{ cm peak}$   
 & start of  $Y_2$  is  $-2.5 \text{ cm}$  at left of screen

$\therefore$  # peak to peak value of  $V_1 = 3 \times 2 \times 10 = 60 \text{ mV}$

& # P-P value of  $V_2 = 3.8 \times 2 \times .05 = 0.38 \text{ V} = 380 \text{ mV}$

& # Period of both signals =  $72 \text{ ms}$

& # frequency of both signals =  $14 \text{ Hz}$

& # Trig. level =  $-2.5 \times .05 = -.125 \text{ V} = -125 \text{ mV}$

& # Trig Slope = positive (because leading up)

& # phase shift =  $\frac{2.4}{7.2} \times 360 = 120^\circ$ ,  $V_1$  lags  $V_2$

# for xy mode,  $X_{pp} = \frac{V_1}{10 \text{ m}} = \frac{60 \text{ m}}{10 \text{ m}} = 6 \text{ cm} = X_m$

&  $Y_{pp} = \frac{V_2}{.5} = \frac{380 \text{ m}}{.5} = 0.76 \text{ cm} = Y_m$

&  $X_0 = X_m \sin \phi = 6 \times \sin 120^\circ = \frac{6\sqrt{3}}{2} = 5.2 \text{ cm}$

&  $Y_0 = Y_m \sin \phi = 0.76 \times \sin 120^\circ = 0.66 \text{ cm}$



$$\boxed{2-14} \quad \int_x n_x = \int_y n_y$$

$$\therefore \int_x * 2 = \frac{\omega}{2\pi} * 1 \Rightarrow \int_x = \frac{\omega}{4\pi} \Rightarrow \omega_x = 2\pi \int_x = \frac{\omega}{2}$$

$\therefore$  Centre of screen is  $(0,0)$  and is a point in the figure.

$$\therefore \omega t_0 + \phi = \frac{\omega t_0}{2} + \bar{\phi} = 0 \therefore \bar{\phi} = -\frac{\omega t_0}{2} = -\frac{1}{2}(-\phi) = \phi/2$$

(OR  $\omega t_0 + \phi = \pi \therefore \frac{\omega t_0}{2} + \bar{\phi} = \pi \therefore \bar{\phi} = \pi - \frac{\omega t_0}{2} = \pi - \frac{1}{2}(-\phi) = \pi + \phi/2$ )

$$\therefore X(t) = 5 \sin\left(\frac{\omega t + \phi}{2}\right) \text{ Cm} \quad \left[ \text{OR } X(t) = 5 \sin\left(\frac{\omega t + \phi + 2\pi}{2}\right) \right]$$

$$\neq Y(t) = 4 \sin(\omega t + \phi) \text{ Cm}$$

a)  $\therefore Y\text{-sensitivity} = \frac{16V}{4C} = 4V/C$

b)  $\neq V_x(t) = 25 \sin\left(\frac{\omega t + \phi}{2}\right) \text{ Volts.}$

c)  $\therefore \omega = 40\pi \text{ rad/ms}$

$$\therefore f = \frac{\omega}{2\pi} = \frac{40\pi}{2\pi \text{ ms}} = 20 \text{ KHz}$$

$$\therefore T = \frac{1}{20\text{k}} = 50 \mu\text{sec}$$

$$\therefore \text{Span of cycle} = \frac{50 \mu\text{s}}{5 \text{ ms/C}} = 10 \text{ C}$$

$\therefore$  One cycle of  $y$  would appear

$\neq$  half a cycle of  $x$  would appear.

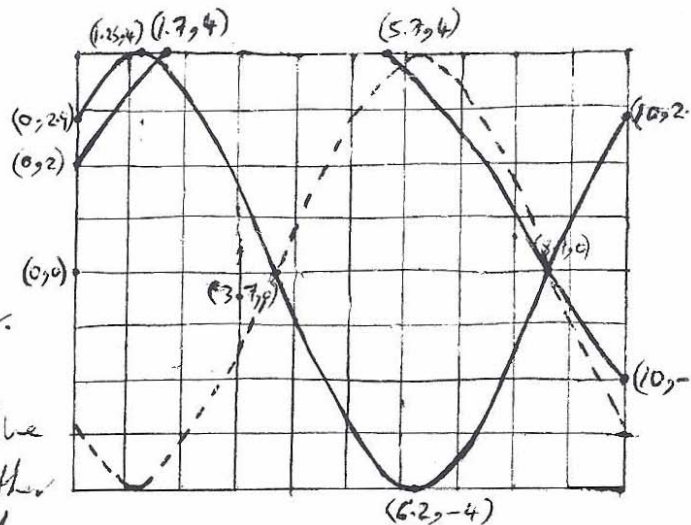
The screen will start at

$X = 2 \text{ Cm}$   $\neq$  slope positive

as shown here (note: the other

possibility is also shown by the

dashed line).



$$\boxed{2-15} \quad n_x \cdot f_x = n_y \cdot f_y$$

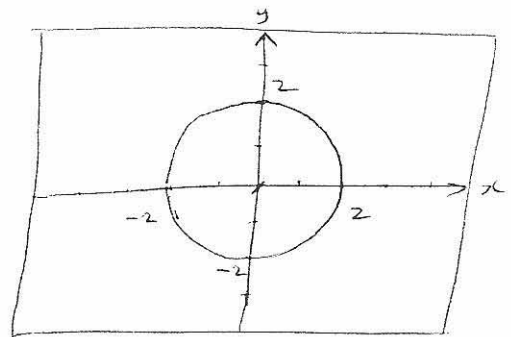
$$\therefore (3\frac{1}{2}) * 50 = 1 * f_y \quad \therefore f_y = \frac{7}{2} * 50 = 175 \text{ Hz}$$

$$\boxed{2-16} \quad X = \frac{V_x}{S_x} = \frac{2 \sin 100\pi t}{1} = 2 \sin 100\pi t \text{ cm}$$

$$Y = \frac{V_y}{S_y} = \frac{4 \cos 100\pi t}{2} = 2 \cos 100\pi t \text{ cm}$$

$$\therefore X^2 + Y^2 = 4 \quad \therefore \text{Circle radius } 2 \text{ cm}$$

a)  $\therefore$  The screen will look like this:



$$b) \quad \left. \begin{aligned} Y_1 &= 2 \sin 100\pi t \\ Y_2 &= 2 \cos 100\pi t \end{aligned} \right\} \begin{aligned} f_1 &= f_2 = \frac{100\pi}{2\pi} = 50 \text{ Hz} \\ \therefore T_1 &= T_2 = \frac{1}{50} = 20 \text{ msec} \end{aligned}$$

$$\therefore \text{Trig. level} = 2 \text{ volts} = \frac{2}{2} = 1 \text{ cm}$$

$$\text{I)} \quad T_1 = T_2 = 20 \text{ msec} = \frac{20 \text{ msec}}{1 \text{ msec/cm}} = \frac{20 \text{ m}}{\text{cm}} = 20,000 \text{ cm}$$

$\therefore$  One period would be seen within 20,000 cm.

$\therefore$  for 10 cm wide screen the change is not visible

$$\therefore Y_2 \Big|_{\text{start of screen}} = 1 \text{ cm, } 100\pi t \Big|_{\text{start of screen}} = \left[ \cos^{-1} \left( \frac{1}{2} \right) \right] = -\pi/3 \quad \& \quad Y_1 \Big|_{\text{start of screen}} = 2 \sin(-\pi/3) = -\sqrt{3} \text{ cm} = -1.732 \text{ cm} \approx -1.7 \text{ cm}$$

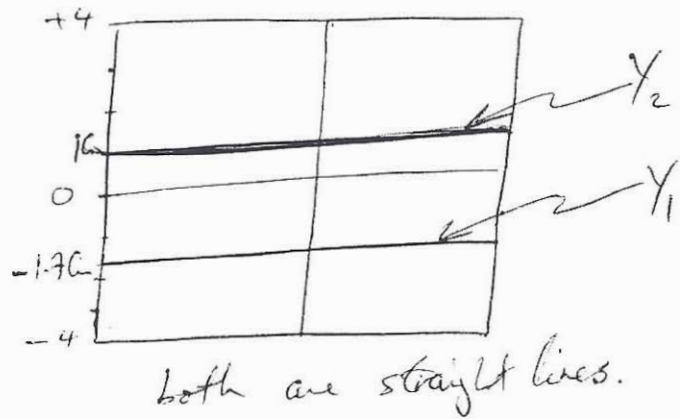
$$Y_2 \Big|_{\substack{\text{end} \\ \text{of screen}}} = 2 \cos 100\pi \left( t_{\substack{\text{start} \\ \text{of screen}}} + 10 \text{ cm} \right) = 2 \cos 100\pi \left( t_{\substack{\text{start} \\ \text{of screen}}} + 10 \times 10^{-3} \right)$$

$$= 2 \cos \left( 100\pi t_{\substack{\text{start} \\ \text{of screen}}} + 100\pi \times 10 \times 10^{-6} \right) = 2 \cos \left( -\frac{\pi}{3} + 10^{-3}\pi \right)$$

$$= 2 \cos \left( -\frac{1}{3} + 0.001 \right) \pi = 2 * 0.503 = 1.01 \text{ cm}$$

$$Y_1 \Big|_{\substack{\text{end} \\ \text{of screen}}} = 2 \sin \left( -\frac{1}{3} + 0.001 \right) \pi = -1.729 \text{ cm}$$

∴ The screen looks like this

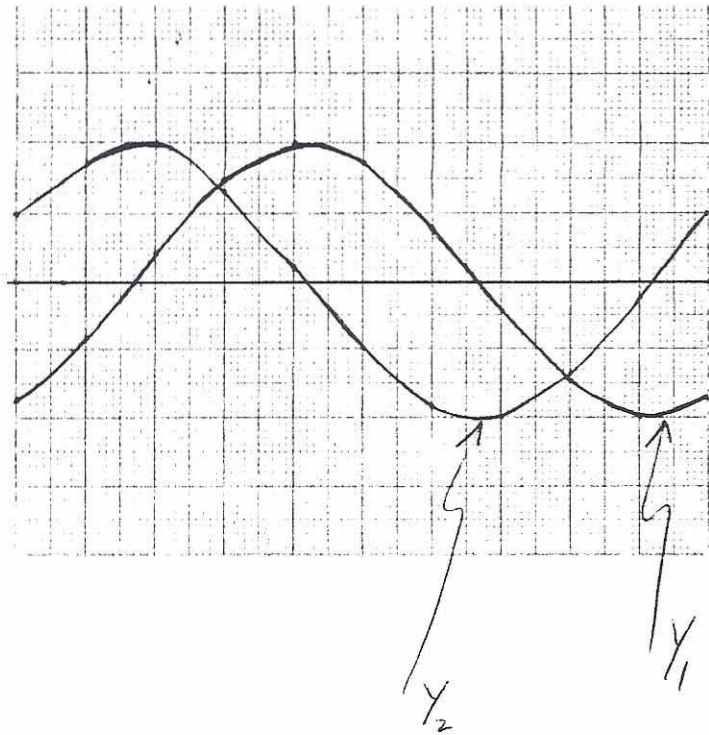


$$\text{II) } T_1 = T_2 = \frac{20 \text{ cm}}{20 \text{ cm/s}} = 10 \text{ cm}$$

∴ One period will take the width of full screen

$Y_2(\text{cm})$	$T(\text{cm})$	$100\pi t$	$Y_1(\text{cm})$
1	0	$-\pi/3$ (start of screen)	-1.73
1.83	1	$-\frac{\pi}{3} + 100\pi * 2 \text{ m} = (-\frac{1}{3} + 2)\pi$	-0.81
1.96	2	$(-\frac{1}{3} + 4)\pi$	+0.42
1.34	3	$(-\frac{1}{3} + 6)\pi$	+1.49
0.21	4	$(-\frac{1}{3} + 8)\pi$	+1.99
-1	5	$(-\frac{1}{3} + 1)\pi$	+1.73
-1.83	6	$(-\frac{1}{3} + 1.2)\pi$	+0.81
-1.96	7	$(-\frac{1}{3} + 1.4)\pi$	-0.42
-1.34	8	$(-\frac{1}{3} + 1.6)\pi$	-1.49
-0.21	9	$(-\frac{1}{3} + 1.8)\pi$	-1.99
1	10	$(-\frac{1}{3} + 2)\pi$	-1.73

Hence, screen looks as follows:



$$\underline{2-17} \text{ @ } V_{1_{rms}} = S_1 * Y_1 / \sqrt{2} = 1 * 2 / \sqrt{2} = \sqrt{2} \text{ Volts} = 1.4142 \text{ Volts}$$

$$dV_{1_{rms}} = S_1 dY_1 / \sqrt{2} = 1 * 1 * 10^{-1} / \sqrt{2} = 0.070711 \text{ Volts} = 70.711 \text{ mV}$$

$$\therefore V_{1_{rms}} = 1.4142 \text{ V} \pm 70.711 \text{ mV} = 1.4142 \text{ V} \pm 5\%$$

$$V_{2_{rms}} = S_2 * Y_2 / \sqrt{2} = 20 \text{ m} * 3 / \sqrt{2} = 42.426 \text{ mV}$$

$$dV_{2_{rms}} = S_2 dY_2 / \sqrt{2} = 20 \text{ m} * 1 * 10^{-1} / \sqrt{2} = 1.4142 \text{ mV}$$

$$\therefore V_{2_{rms}} = 42.426 \text{ mV} \pm 1.4142 \text{ mV} = 42.426 \text{ mV} \pm 3.33\%$$

$$\textcircled{b} f_1 = f_2 = f \quad T_1 = T_2 = T \quad \text{with } fT = 1$$

$$\therefore T = 8 * 20 \text{ m} = 160 \text{ msec}$$



$$\therefore f = \frac{1}{160\text{m}} = 6.25 \text{ Hz}$$

$$\therefore f = \frac{1}{T} \quad \therefore \frac{\Delta f}{f} = -\frac{\Delta T}{T}$$

$$\therefore \% \text{ error in } f = \% \text{ error in } T = \frac{1}{8} * 100 = 1.25 \% = 7.8125 * 10^{-2}$$

$$\therefore f = f_1 = f_2 = 6.25 \text{ Hz} \pm 0.078125 \text{ Hz} = 6.25 \text{ Hz} \pm 1.25 \%$$

(c) Trig level = 0 Volts  
slope positive.

$$(d) \frac{\phi}{180} = \frac{2}{4} = \frac{1}{2} \quad \therefore \phi = 90^\circ$$

$\therefore V_2$  leads  $V_1$  by  $90^\circ$  or  $V_1$  lags  $V_2$  by  $90^\circ$

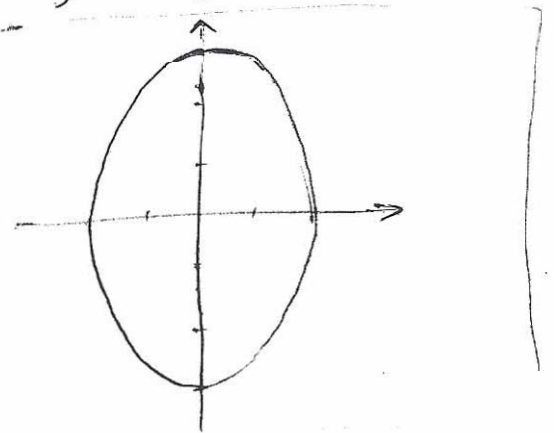
(e) The screen at xy mode is given by

$$Y = Y_2 = 3 \cos \omega t$$

$$X = Y_1 = 2 \sin \omega t$$

$$\therefore \left(\frac{Y}{3}\right)^2 + \left(\frac{X}{2}\right)^2 = 1$$

$\therefore$  ellipse as shown



2-18  $n_x f_x = n_y f_y \Rightarrow \therefore n_x = \frac{1}{2}, n_y = 1\frac{1}{2}, f_x = 30 \text{ Hz}$

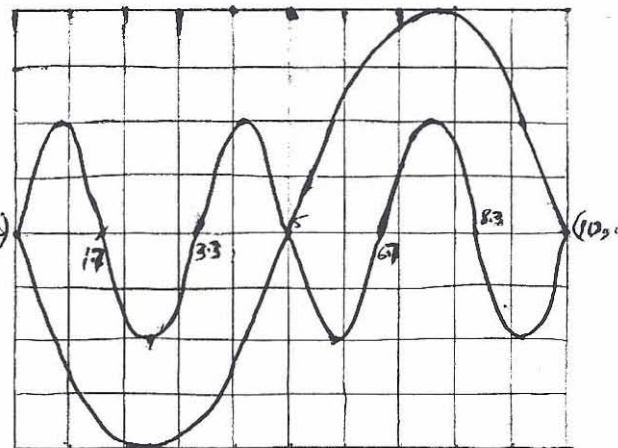
$$\therefore f_y = \frac{1}{2} * 30 / 1\frac{1}{2} = 10 \text{ Hz}$$

Since centre of screen (0,0) is one of the figure points and positive peaks form a corner,

$$\therefore Y = -4 \sin 20\pi t \text{ cm.}$$

$$T_y = \frac{1}{10} \text{ sec} = \frac{100 \text{ ms}}{10 \text{ ms/cm}} = 10 \text{ cm}$$

$\therefore$  A screen will display one period of  $y$  + 3 off  $x$  as shown above





2-19  $Y_1(t) = \frac{15}{5} \sin 50\pi t = 3 \sin 50\pi t \text{ Cm}$

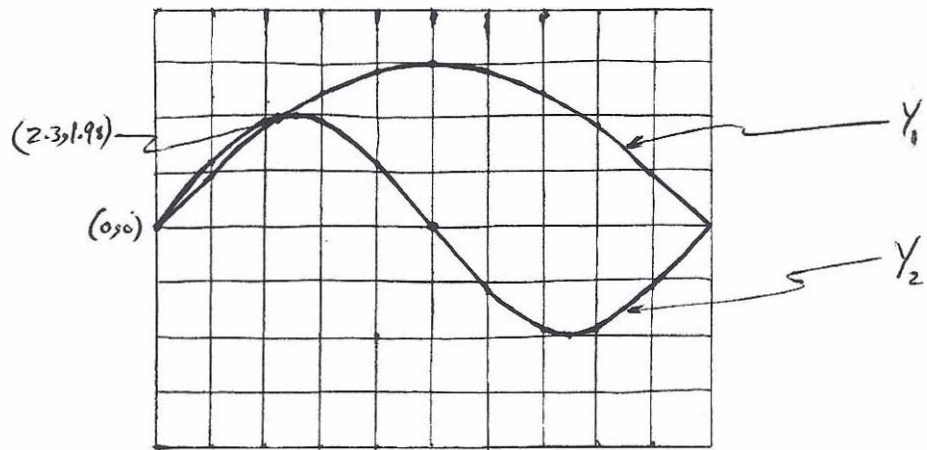
$\& Y_2(t) = \frac{20}{10} \sin 100\pi t = 2 \sin 100\pi t \text{ Cm}$

a)  $f_1 = \frac{50\pi}{2\pi} = 25 \text{ Hz} \Rightarrow T_1 = 40 \text{ ms} = 20 \text{ Cm}$

$\& f_2 = \frac{100\pi}{2\pi} = 50 \text{ Hz} \Rightarrow T_2 = 20 \text{ ms} = 10 \text{ Cm}$

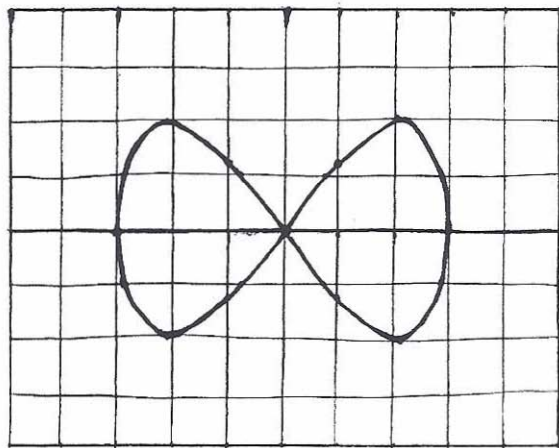
$\therefore$  Screen will show one cycle of  $Y_2$   $\&$  half cycle of  $Y_1$ , as below

F



b)  $X = Y_1 = 3 \sin 50\pi t \quad \& \quad Y = Y_2 = 2 \sin 100\pi t$

$\therefore Y = 2 \sin(2 * 50\pi t) = 2 \sin(2 * \sin^{-1} \frac{X}{3}) = 4 \cdot \sin(\sin^{-1} \frac{X}{3}) \cos(\sin^{-1} \frac{X}{3})$   
 $= 4 \cdot \frac{X}{3} \sqrt{1 - (\frac{X}{3})^2} = \frac{4}{9} X \sqrt{9 - X^2}$  symmetric about X-Y axis.  
 The shape looks like  $\infty$  and is plotted point by point as shown below



2-20

$$V_1 = (3 \text{ cm}) (50 \text{ mV/cm}) = 150 \text{ mV}$$

$$V_2 = (2 \text{ cm}) (50 \text{ mV/cm}) = 100 \text{ mV}$$

$$T = (6 \text{ cm}) (20 \text{ m/s/cm}) = 120 \text{ ms} = 0.12 \text{ s}$$

$$\omega = \frac{2\pi}{T} = \frac{2\pi}{0.12} = 52.359878 \text{ rad/s} \approx 52.36 \text{ rad/s}$$

$$\phi = \phi_2 - \phi_1 = -\frac{1}{6} 360^\circ = -60^\circ$$

3-5 a) No. of digits are  $6\frac{1}{2}$

b)  $f = \pi \text{ Hz} = 3.1415927 \text{ Hz}$

$\& T = 0.31830989 \text{ sec}$

The best reading is the one giving least getting error, i.e. max. possible no. of digits with high significance.

- i) If LSD is  $1\mu\text{s}$  screen is 318309 with %e  $\leq \frac{1}{318309} \times 100$
- ii) If LSD is  $10\mu\text{s}$  = = 031830 = =  $\leq \frac{1}{31830} \times 100$
- iii) other period ranges and are seen to be all, but the first, inaccurate
- iv) If LSD is  $10\text{Hz}$  screen is 000000 with %e = 100%
- v) other frequency ranges are going also to give 000000  $\& \%e = 100\%$

c)  $\therefore$  The best reading is obtained by pressing  $1\mu\text{s}$  for LSD and is 318309  $\mu\text{s}$  i.e.  $f = 3.1416014 \text{ Hz}$

d) Uncertainty =  $\frac{1}{318309} \times 100 = 0.000314\%$  in both T  $\& f$   
(note actual error =  $2.8 \times 10^{-4}\%$  < uncertainty  $\therefore$  OK)

e) period ranges are 1999999 (10ms)  $\&$  000001 (1 $\mu\text{s}$ )  
i.e.  $T \in (2 \times 10^6 \times 10\text{ms}, 1\mu\text{s}]$  i.e.  $f \in (5 \times 10^{-5}, 10^6] \text{ Hz}$

$\&$  frequency ranges are 000001 (10Hz)  $\&$  1999999 (100kHz)  
i.e.  $f \in [10, 2 \times 10^6 \times 100 \times 10^3) \text{ Hz} = [10, 2 \times 10^{11}) \text{ Hz}$

$\therefore$  The frequencies measurable by this DFC are  $(5 \times 10^{-5}, 10^6] \cup [10, 2 \times 10^{11}) \text{ Hz}$   
i.e.  $(5 \times 10^{-5}, 2 \times 10^{11}) \text{ Hz}$

$$\boxed{3-6} \quad V_o(\text{true}) = 8 * \frac{15}{80+15} = 1.2631579 \text{ volts}$$

$$\text{I) } 2\text{mV range} \quad \therefore R_m = \frac{1\text{K}\Omega}{V_{FS}} * 2\text{mV} = 2\Omega$$

$$\therefore V_o = 8 * \frac{15 \parallel 2}{15 \parallel 2 + 80} = 1.726618705 = 172.6618705 \text{ mV (too much loading)}$$

$$\therefore \text{Reading is } 1.999999 \text{ mV (overflow)} \neq \%e = 99.84\%$$

$$\text{II) } 0.2\text{V range} \quad \therefore R_m = 1\text{K} * .2 = .2\text{K} = 200\Omega$$

$$\therefore V_o = 8 * \frac{15 \parallel 200}{15 \parallel 200 + 80} = 1.18811881188 \text{ volts (little loading)}$$

$$\therefore \text{Reading is } .1999999 \text{ volts (overflow)} \neq \%e = 84.17\%$$

$$\text{III) } 20\text{V range} \quad \therefore R_m = 1\text{K} * 20 = 20\text{K}\Omega = 20,000\Omega$$

$$\therefore V_o = 8 * \frac{15 \parallel 20,000}{15 \parallel 20,000 + 80} = 1.26236061434 \text{ volts (less loading)}$$

$$\therefore \text{Reading is } 1.26236 \text{ volts} \neq \%e = 0.0632\%$$

$$\text{IV) } 2000\text{V range} \quad \therefore R_m = 1\text{K} * 2000 = 2\text{M}\Omega = 2,000,000\Omega$$

$$\therefore V_o = 8 * \frac{15 \parallel 2,000,000}{15 \parallel 2,000,000 + 80} = 1.26314991694 \text{ volts (least loading)}$$

$$\therefore \text{Reading is } 001.263 \text{ volts} \neq \%e = 0.0125\%$$

$\therefore$  a) best range is the 2KV range, giving:

$\neq$  b) The best reading of 001.263 Volts with:

$\neq$  c) The least error of 0.0125%.

3-7

$$f = 60 \text{ Hz}$$

$$T = 16.6\bar{6} \text{ ms}$$

i)	If LSD is 10ms	screen is	00001	with %e $\leq \frac{1}{1} \times 100$
ii)	" = " 1sec	" = "	00000	" = " = 100%
iii)	" = " 100s	" = "	"	" = " = "
iv)	" = " 100 Hz	" = "	"	" = " = "
v)	" = " 10 Hz	" = "	00006	" = " $\leq \frac{1}{6} \times 100$
vi)	" = " 1 Hz	" = "	00060	" = " $\leq \frac{1}{60} \times 100$

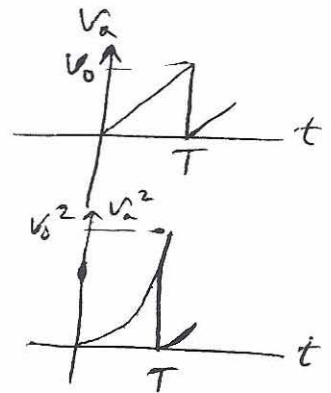
$\therefore$  The least uncertain range is the 1 Hz range for LSD

and best reading is 00060 Hz with actual error %  
(note  $0\% \leq \frac{100\%}{60} = 1.7\% \therefore \text{OK}$ )



4-8 # waveform (a)

$$V_{a_{rms}} = \sqrt{\frac{1}{T} * \frac{1}{3} * V_0^2 * T} = \frac{V_0}{\sqrt{3}}$$



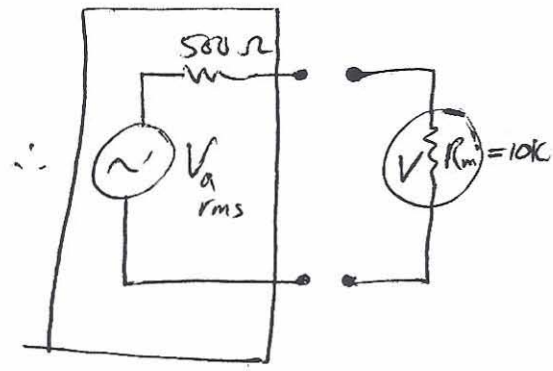
∴ Reading due to waveform  
(assuming no loading) =

$$= 1.11 * \text{Av. of FWR signal} =$$

$$= 1.11 * \left(\frac{1}{T} * \frac{V_0 * T}{2}\right) = \frac{1.11 V_0}{2}$$

$$\therefore \text{Waveform error} = \frac{\frac{V_0}{\sqrt{3}} - \frac{1.11 V_0}{2}}{\frac{V_0}{\sqrt{3}}}$$

$$= \frac{2 - 1.11\sqrt{3}}{2} = 0.0387 = 3.87\%$$



f Reading due to loading (assuming no waveform error) =

$$= \frac{10K}{10K + 500} * V_{a_{rms}} = 0.952 V_{a_{rms}}$$

$$\therefore \text{Loading error} = \frac{V_{a_{rms}} - 0.952 V_{a_{rms}}}{V_{a_{rms}}} = 0.0476 = 4.76\%$$

f Reading due to both waveform & loading =

$$= 1.11 * \text{Av. of FWR signal} = 1.11 * \left(\frac{1}{T} * \frac{10K V_0}{500 + 10K} * T/2\right) =$$

$$= \frac{1.11}{2} * 0.952 V_0 = 0.528 V_0$$

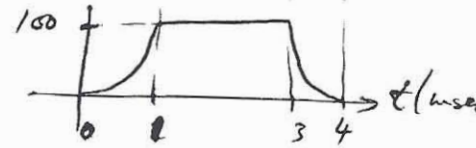
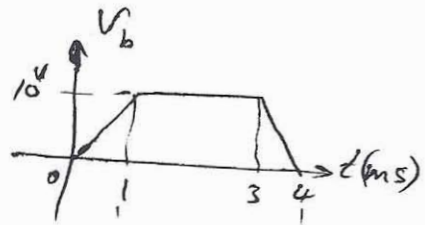
$$\therefore \text{Total error} = \frac{\frac{V_0}{\sqrt{3}} - 0.528 V_0}{\frac{V_0}{\sqrt{3}}} = 0.085 = 8.5\%$$

$$\left(\text{∴ waveform error} + \text{loading error} = 3.87 + 4.76 = 8.6\%\right)$$

# waveform (b)

$$V_{b_{rms}} = \sqrt{\frac{1}{4} \left( \frac{100 \times 1}{3} + 100 \times 2 + \frac{100 \times 1}{3} \right)}$$

$$= \sqrt{\frac{1}{4} \left( \frac{200 + 600}{3} \right)} = \sqrt{\frac{800}{12}} = \frac{20}{\sqrt{6}} \text{ V}$$



$\therefore$  Reading due to waveform (no loading) =

$$= 1.11 * \text{Av. of FWR signal} = 1.11 * \left( \frac{1}{4} * \left( \frac{10 \times 1}{2} + 10 \times 2 + \frac{10 \times 1}{2} \right) \right)$$

$$= 1.11 * \frac{30}{4} = 8.33 \text{ V}$$

$$\therefore \text{Waveform error} = \left| 1 - \frac{8.33}{20/\sqrt{6}} \right| = | -0.0196 | = 1.96 \%$$

$\&$  Reading due to loading is as before (assuming no waveform error) = 4.7

$\&$  Reading due to both waveform  $\&$  loading =

$$= 1.11 * \text{Av. of FWR signal} = 1.11 * \left( 0.952 * \frac{30}{4} \right) =$$

$$= 7.93 \text{ V}$$

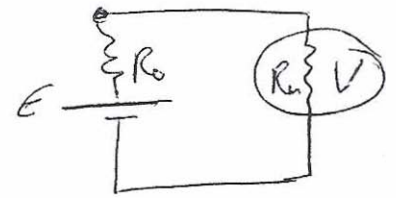
$$\therefore \text{Total error} = 1 - \frac{7.93}{20/\sqrt{6}} = 0.0293 = 2.93 \%$$

$$\left( \bar{n} \text{ waveform error} + \text{loading error} = -1.96 + 4.76 = 2.8 \% \right)$$

$$\boxed{4-10} \# \% \text{ loading error} = \left(1 - \frac{\text{Reading}}{\text{True Value}}\right) \times 100$$

$$= \left(1 - \frac{E \cdot R_m}{R_m + R_o}\right) \times 100 = \frac{R_o}{R_m + R_o} \times 100$$

$$= \frac{400}{400 + 10 \times 10^6} \times 100 = 0.004 \% \text{ (almost nothing)}$$



$$\therefore \# TV = AV_{DVM} = \frac{\sum \text{DVM readings}}{4} = 3.412 \text{ Volts}$$

$$AV_{MCV} = \frac{\sum \text{MCV readings}}{5} = 3.22 \text{ Volts}$$

$$\therefore \text{Precision of MCV} = |3.22 - 3.24| = 0.02 \text{ Volts}$$

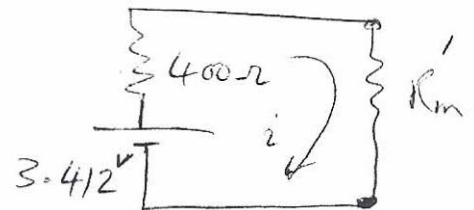
$$\# \text{ Precision of DVM} = |3.412 - 3.413| = 0.001 \text{ Volts} = \text{Accuracy of DVM}$$

$$\# \text{ Accuracy of MCV} = |3.412 - 3.20| = 0.212 \text{ Volts}$$

# Only MCV readings need recalibration by adding

$$\text{Bias} = 3.412 - 3.22 = 0.192 \text{ Volts}$$

# Bias is due to drop  
at the \$400 \Omega\$ battery  
resistance



$$i = \frac{0.192}{400} = .48 \text{ mA}$$

$$R_m' = \frac{3.412}{.48 \times 10^{-3}} - 400 = 6.708 \text{ K}\Omega, \text{ I/p resistance of MCV}$$

$$\boxed{4-11} \# T_{FSD} = N B A i_{FSD}^2 = 100 * .1 * 5 * (10^{-2})^2 * \frac{V_{FSD}}{R_m} =$$

$$= 5 * 10^{-3} * \frac{100 * 10^{-3}}{R_m} = \frac{5 * 10^{-4}}{R_m}$$

$$\text{but } T_{FSD} = K O_{FSD} = 5 * 10^{-6}$$

$$\therefore 5 * 10^{-6} = \frac{5 * 10^{-4}}{R_m} \quad \therefore R_m = 100 \Omega$$

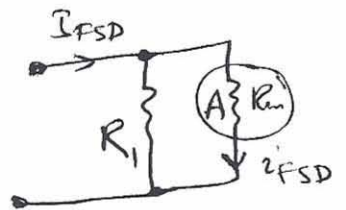
$$\therefore i_{FSD} = \frac{100 \text{ m}}{100} = 1 \text{ mA}$$

$$\# I_{FSD} = 50 \text{ mA}$$

$$\therefore \frac{i_{FSD}}{I_{FSD}} = \frac{R_1}{R_1 + R_m}$$

$$\therefore \frac{50 \text{ m}}{1 \text{ m}} = 1 + \frac{R_m}{R_1} \Rightarrow 50 = 1 + \frac{R_m}{R_1} \quad \therefore \frac{R_m}{R_1} = 49$$

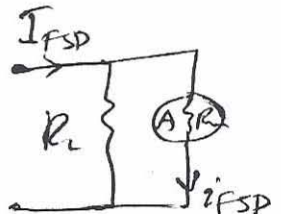
$$\therefore \text{Choose } R_1 = \frac{R_m}{49} = \frac{100}{49} = 2.041 \Omega \text{ shunt.}$$



$$\# I_{FSD} = 1 \text{ Amp}$$

$$\therefore \frac{1}{1 \text{ m}} = 1 + \frac{R_m}{R_2} \quad \therefore \frac{R_m}{R_2} = 1000 - 1 = 999$$

$$\therefore \text{Choose } R_2 \text{ to be } \frac{R_m}{999} = \frac{100}{999} = 0.1 \Omega \text{ shunt}$$

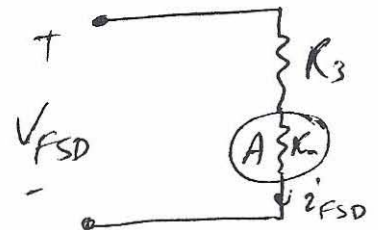


$$\# V_{FSD} = 10 \text{ V}$$

$$\therefore R_3 = \frac{V_{FSD}}{i_{FSD}} - R_m = \frac{10}{1 \text{ m}} - 100$$

$$= 10 \text{ K} - 100 = 9.9 \text{ K} \Omega$$

$$\therefore \text{Choose } R_3 \text{ to be } 9.9 \text{ K} \Omega \text{ Series}$$



$$\# V_{FSD} = 200 \text{ V}$$

$$\therefore \text{choose } R_4 \text{ to be } \frac{200}{1 \text{ m}} - 100 = 200 \text{ K} - 1 \text{ K} = 199.9 \text{ K} \Omega \text{ Series}$$



$$\boxed{4-19} \quad T = BANi = mg \cdot C_g \cdot \sin \theta$$

$$\therefore i = \frac{m \cdot g \cdot C_g \cdot \sin \theta}{BAN} = \frac{10 \times 10^{-3} \times 9.81 \times 2 \times 10^{-2} \cdot \sin \theta}{0.3 \times 4 \times (10^{-2})^2 \times 100}$$

$$= 0.1635 \sin \theta \text{ Amps}$$



$$\therefore \text{for } \theta = 30^\circ \Rightarrow i = 81.75 \text{ mA}$$

$$\& \text{ for } \theta = 60^\circ \Rightarrow i = 141.6 \text{ mA}$$

$$\& \text{ for } \theta = 90^\circ \Rightarrow i = i_{\text{FSD}} = 163.5 \text{ mA}$$

If  $i$  was increased beyond that no restoring torque can stop movement since the heaviest gravitational torque occurs at  $\theta = 90^\circ$ .

$$\boxed{4-20} \quad \text{rms}^2 = \frac{1}{T} \cdot \left[ \frac{1}{3} \times \frac{T}{4} \times E^2 + E^2 \times \frac{3T}{4} \right] = E^2 \cdot \left( \frac{1}{12} + \frac{3}{4} \right) = \frac{E^2}{12} \times 10$$

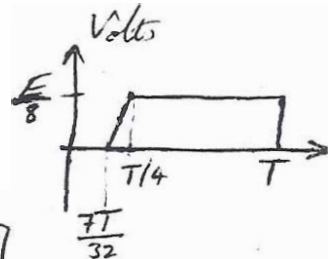
$$\therefore \text{rms} = \sqrt{\frac{10}{12}} E$$

$$\text{av.} = \frac{1}{T} \cdot \left[ \frac{1}{2} \times \frac{T}{4} \times E + E \times \frac{3T}{4} \right] = E \cdot \left( \frac{1}{8} + \frac{3}{4} \right) = \frac{7E}{8} = \text{dc}$$

$$\therefore \text{ac} = \sqrt{\text{rms}^2 - \text{dc}^2} = \sqrt{\frac{10E^2}{12} - \frac{49E^2}{64}} = E \sqrt{\frac{160 - 147}{192}} = E \sqrt{\frac{13}{192}}$$

Scale factor of HW-AC-MCV is  $\frac{\pi}{\sqrt{2}}$

The signal input to the movement is as shown:



$$\therefore \text{Reading} = \frac{\pi}{\sqrt{2}} \times \text{Av. of it} = \frac{\pi}{\sqrt{2}} \times \frac{1}{T} \left[ \frac{E}{8} \times \frac{T}{32} \times \frac{1}{2} + \frac{E}{8} \times \frac{3T}{4} \right]$$

$$= \frac{\pi}{\sqrt{2}} \cdot E \cdot \left( \frac{1 + 48}{512} \right) = \frac{49\pi}{512\sqrt{2}} \cdot E$$

$$\therefore \%e = \left| 1 - \frac{49\pi E / 512\sqrt{2}}{E \sqrt{13} / \sqrt{192}} \right| \times 100 = 18.3\%$$



$$\boxed{4-21} \text{ a) } 100\mu = 10 * \frac{R_1}{R_1 + 20}$$

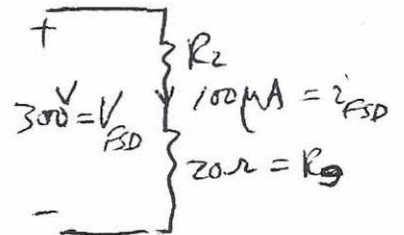
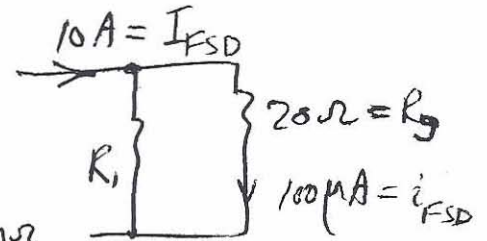
$$\therefore \frac{R_1 + 20}{R_1} = \frac{10}{100\mu} = 100 \text{ K}$$

$$\therefore 1 + \frac{20}{R_1} = 100 \text{ K} \Rightarrow R_1 = 0.200002 \text{ m}\Omega$$

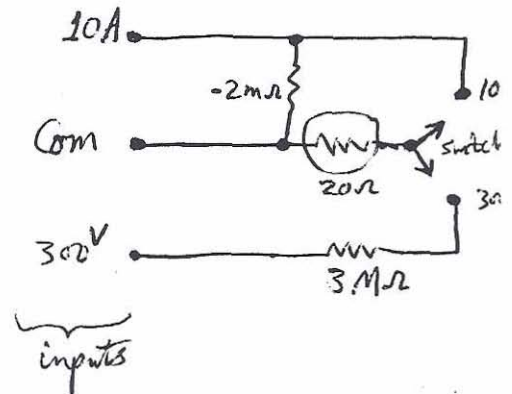
\(\therefore\) We use a shunt of  $0.200002 \text{ m}\Omega$

$$\text{f) } \frac{300}{100\mu} = R_2 + 20 \Rightarrow R_2 = 2999980 \Omega$$

\(\therefore\) We use a series of  $\approx 3 \text{ M}\Omega$



To arrange for both combination the following diagram is followed, where the current input is between Com & 10 terminals with switch on 10 Amp & the voltage input is between Com & 300V terminals with switch on 300V position.



$$\text{b) } i_{FSD} = I_{FSD} * \frac{R_1}{R_1 + R_g} \Rightarrow I_{FSD} = \left( \frac{R_1 + R_g}{R_1} \right) \cdot i_{FSD} = \left( 1 + \frac{R_g}{R_1} \right) i_{FSD}$$

$$\therefore \Delta I_{FSD} = \frac{i_{FSD}}{R_1} \Delta R_g + \left( 1 + \frac{R_g}{R_1} \right) \Delta i_{FSD} - \frac{i_{FSD} \cdot R_g}{R_1^2} \Delta R_1$$

$$= \frac{R_g}{R_1} \cdot i_{FSD} \left( \frac{\Delta R_g}{R_g} - \frac{\Delta R_1}{R_1} \right) + \left( 1 + \frac{R_g}{R_1} \right) \Delta i_{FSD}$$

$$\therefore \Delta I_{FSD} = \sqrt{\left( \frac{20}{0.2\text{m}} * 100\mu \right)^2 \cdot (-0.02 + 0.1)^2 + \left( 1 + \frac{20}{0.2\text{m}} \right)^2 (1\mu)^2} = 1.0247 \text{ Amp} = 10.25\%$$

$$\text{f) } V_{FSD} = i_{FSD} * (R_g + R_2) \quad \therefore \Delta V_{FSD} = (R_g + R_2) \Delta i_{FSD} + i_{FSD} (\Delta R_g + \Delta R_2)$$

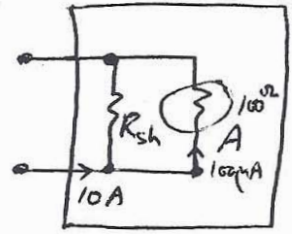
$$\therefore \Delta V_{FSD} = \sqrt{(3\text{M})^2 * (1\mu)^2 + (100\mu)^2 \cdot [(0.02 * 20)^2 + (0.1 * 3\text{M})^2]} = 30.15 \text{ Volts} = 10.05\%$$

4-22  $R_{sh} (10 - 100 \times 10^{-6}) = 100 \times 100 \times 10^{-6}$

$\therefore R_{sh} = 1.00001 \text{ m}\Omega$

$\therefore$  We can obtain a 10 Amp range

by shunting the movement with about  $1 \text{ m}\Omega$  resistor.



new Ammeter

The new ammeter will now have 10 Amp FSD & an internal resistance  $R_g = R_{sh} \parallel 100 = 1 \text{ m}\Omega$

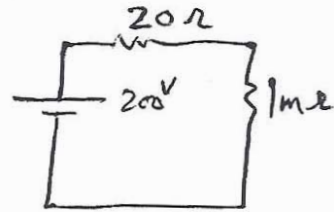
a) The reading =  $\frac{200}{20 + 1 \text{m}} = 9.9995 \text{ Amps}$

b) True value =  $\frac{200}{20} = 10 \text{ Amps}$

$\therefore$  Bias =  $10 - 9.9995 = .5 \text{ mAmp}$

$\% \text{ e} = 0.005 \%$

This error is caused by loading (little loading).



c) i) 20 A range  $\therefore R_m = \frac{1}{50 \text{mV} \times 20} = 1 \Omega \quad \therefore i = \frac{200}{20 + 1} = 9.52381$

$\therefore$  Reading is 9.52 Amp  $\% \text{ e} = 4.8\%$  (too much loading killed)

ii) 200 A range  $\therefore R_m = \frac{1}{50 \text{mV} \times 200} = 0.1 \Omega \quad \therefore i = \frac{200}{20 + 0.1} = 9.95025$

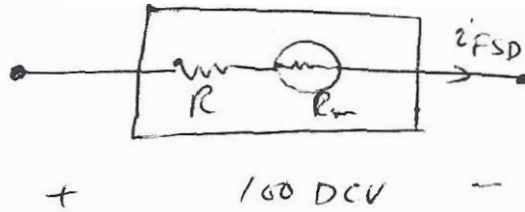
$\therefore$  Reading is 09.9 Amp  $\% \text{ e} = 1\%$  (less loading high gain error)

iii) 2 KA range  $\therefore R_m = \frac{1}{50 \text{mV} \times 2 \text{K}} = 10 \text{ m}\Omega \quad \therefore i = \frac{200}{20 + 0.01} = 9.99500 \text{ A}$

$\therefore$  Reading is 009.9 Amp  $\% \text{ e} = 10\%$  (least loading but highest gain error)

$\therefore$  If DAM is used, the best range is the 200 Amp range since it gives least overall error of 1%.

4-23



$$\therefore \frac{100}{2FSD} = R + R_m$$

$$\therefore R = \frac{100}{100\mu} - 100 = 1M - 100\Omega = 999.9K\Omega$$

#  $\therefore$  We connect  $999.9K\Omega$  series resistor to get 100 DCV voltmeter, hence  $V$  has  $1M\Omega$  as  $R_m$  with 100 DCV <sub>FSD</sub>

$$(a) \text{ Reading} = 200 * \frac{(10K || 1M)}{20K + (10K || 1M)} = 66.2252 \text{ Volts}$$

$$(b) \text{ bias} = \text{True Value} - AV = 200 * \frac{10}{10+20} - 66.2252 = 0.4415 \text{ Volts}$$

$$\# \%e = \frac{\text{bias}}{TV} = \frac{0.4415}{200/3} * 100 = 0.66\%$$

$$(c) \# 60V \text{ range} \quad \therefore R_{m1} = 1K * 60 = 60K\Omega$$

$$\therefore \text{voltage}_1 = 200 * \frac{10 || 60}{20 + 10 || 60} = 59.9999 \text{ volts}$$

$$\therefore \text{Reading}_1 = 59.9 \text{ Volts} \quad \neq \%e = 10.0\%$$

$$\# 70V \text{ range} \quad \therefore R_{m2} = 1K * 70 = 70K$$

$$\therefore \text{voltage}_2 = 200 * \frac{10 || 70}{20 + 10 || 70} = 60.8696 \text{ volts}$$

$$\therefore \text{Reading}_2 = 60.8 \text{ volts} \quad \neq \%e = 8.7\%$$

$$\# 80V \text{ range} \quad \therefore R_{m3} = 1K * 80 = 80K$$

$$\therefore \text{voltage}_3 = 200 * \frac{10 || 80}{20 + 10 || 80} = 61.5385 \text{ Volts}$$

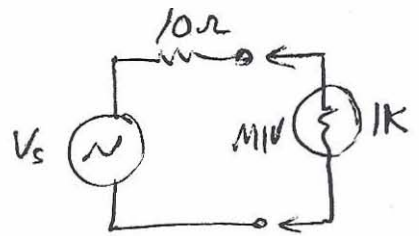
$$\therefore \text{Reading}_3 = 61.5 \text{ Volts} \quad \neq \%e = 7.7\%$$

$\therefore$  we select the 80V range since it gives least error of 7.7%

64

**5-3** The rms as in **4-20** is given by

$$rms = \sqrt{\frac{10}{12}} \cdot E = V_{s_{rms}}$$



$$\therefore \text{reading of MIV} = \left( \frac{1K}{10+1K} \right) V_{s_{rms}}$$

$$= \frac{1000}{1010} * \sqrt{\frac{10}{12}} E = 0.9038 E$$

$$\% e = \left| 1 - \frac{1000 \sqrt{10} E / 1010 \sqrt{12}}{\sqrt{10} E / \sqrt{12}} \right| * 100 = \left| 1 - \frac{1000}{1010} \right| * 100 = \frac{1000}{1010} \approx 1\%$$

$\therefore$  Error is about 1% and it is due to loading of MIV.

**5-4** MIV reads rms straight with no loading, hence it is correct

$$\text{(a) } \therefore V_{rms} = \text{MIV reading} = \sqrt{\left( 100^2 * \frac{I}{3} * \frac{1}{3} + 100^2 * \frac{I}{3} + 0 \right) / T} =$$

$$= 100 \sqrt{\frac{1}{9} + \frac{1}{3}} = 100 \sqrt{\frac{1+3}{9}} = 100 * \frac{2}{3} = \frac{200}{3} = 66.6$$

#  $\therefore$  MIV reading = RMS of waveform = 66.6 volts

# MCV reading = 1.11 \* Av of FWR signal =

$$= 1.11 * \left[ \left( 100 * \frac{I}{3} * \frac{1}{2} + 100 * \frac{I}{3} + 0 \right) / T \right]$$

$$= 1.11 * 100 \left( \frac{1}{6} + \frac{1}{3} \right) = 1.11 * 100 * \frac{1}{2} = 55.54$$

#  $\therefore$  MCV reading = 55.54 volts

(b) MCV has an error of 11.13 Volts = 16.7%

(c) Error in MCV is due to waveform since it is made to measure sinusoidal voltage only not other waveforms



$$\boxed{5-5} \quad \text{RMS}^2 = \frac{1}{T} \left[ 100^2 * \frac{1}{3} * \frac{T}{3} + 100^2 * \frac{T}{3} + 0 \right] = \frac{100^2}{9} (1+3) = \frac{4}{9} * 100^2 \text{ V}^2$$

$$\therefore \text{RMS} = \frac{2 * 100}{3} = \frac{200}{3} = 66.6 \text{ Volts}$$

$$\text{DC} = \frac{1}{T} \left[ 100 * \frac{1}{2} * \frac{T}{3} + 100 * \frac{T}{3} + 0 \right] = \frac{100}{6} (1+2) = 50 \text{ Volts}$$

$$\therefore \text{AC} = \sqrt{66.6^2 - 50^2} = 44.09586 \text{ Volts}$$

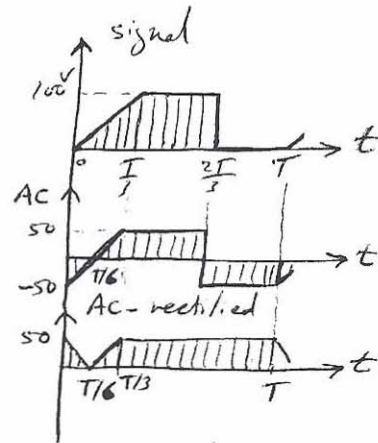
- a) The MIV will read the RMS i.e. 66.6 Volts with 0% error.  
The MCV will read good for AS sinusoids.

$$\therefore \text{MCV reading} = \frac{\pi}{2\sqrt{2}} * (\text{AC-rectified})_{\text{av}}$$

$$= \frac{\pi}{2\sqrt{2}} * \left[ \frac{1}{T} * (50 * T - 50 * \frac{T}{3} * \frac{1}{2}) \right]$$

$$= \frac{\pi}{2\sqrt{2}} * 50 \left( 1 - \frac{1}{6} \right) = \frac{25\pi}{\sqrt{2}} * \frac{5}{6} = \frac{125\pi}{6\sqrt{2}}$$

$$\therefore \text{MCV reading is } 46.2800 \text{ Volts} \approx \text{AC}$$



- b) Since MIV gives RMS & MCV gives approximately AC

$$\therefore \text{DC} = \sqrt{(\text{RMS})^2 - (\text{AC})^2} \approx \sqrt{(\text{MIV})_{\text{reads}}^2 - (\text{MCV})_{\text{reads}}^2} = \sqrt{66.6^2 - 46.28^2} = 47.985$$

- \(\therefore\) My measurement will be the square-root of the squares-difference of both readings = 47.985 Volts

$$\text{c) } \%e = \left| 1 - \frac{47.985}{50} \right| * 100 = 4.03\%$$

This error is caused by MCV and is due to waveform factor

$$\boxed{5-6} \quad \text{AV.} = \frac{4-2}{2} = 1 \text{ Volt} = \text{dc}$$

$$\text{AC} = \frac{4+2}{2\sqrt{3}} = \sqrt{3} \text{ Volts}$$

$$\text{RMS} = \sqrt{\text{AC}^2 + \text{dc}^2} = \sqrt{3+1} = 2 \text{ Volts}$$

- \(\therefore\) MIV reads RMS of 2 Volts & \%e = 0%  
& MCV reads  $1.11 * \frac{3}{2} = 1.666 \text{ Volts}$  & \%e = 16.7% (waveform error for ac+dc is 66)



$$\boxed{6-8} \quad (10 + j\omega L) \left(18 + \frac{1}{j\omega C}\right) = 20 \times 12$$

$$\therefore (10 + j\omega L) (1 + j\omega C(18)) = 20 \times 12 j\omega C$$

$$\therefore 10 - \omega^2 L C (18) = 0 \Rightarrow \omega^2 L C = \frac{10}{18} \quad (1)$$

$$\text{f } 180 \omega C + \omega L = 240 \omega C \Rightarrow \omega L = 60 \omega C \quad \text{OR } L = 60 C \quad (2)$$

$$(2) \text{ into } (1) \quad \therefore \omega^2 (60 C) C = \frac{10}{18}$$

$$\therefore \omega^2 C^2 = \frac{10}{18 \times 60} = \frac{1}{108} \quad \therefore C = \frac{1}{\omega \sqrt{108}} = \frac{1}{2\pi \times 60 \sqrt{108}} = 2.552 \times 10^{-4} \text{ F}$$

$$\therefore C = 255.2 \mu\text{F}$$

$$\text{f } L = 60 C = 15314.7 \mu\text{H} = 15.315 \text{ mH}$$

$L$  is 15.3 mH &  $C$  is 255.2  $\mu\text{F}$

**6-9**

At balance i.e. at  $25^\circ\text{C}$   $R^2 = 4\text{M} \therefore R = 2\text{K}$

Assuming little variation with temperature & taking equivalent at (A) ends:

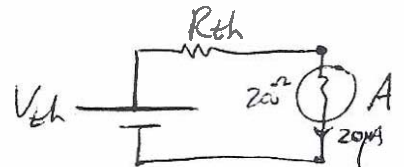
$$\therefore R_{th} = (R \parallel 4\text{K}) + (R \parallel 1\text{K}) \approx 2 \parallel 4 + 2 \parallel 1 = \frac{8}{6} + \frac{2}{3} = 2\text{K}$$

$$\text{f } V_{th} = 12 * \left[ \frac{4}{4+R+\Delta R} - \frac{R+\Delta R}{1+R+\Delta R} \right] = 12 * \frac{4(3+\Delta R) - (2+\Delta R)(6+\Delta R)}{(6+\Delta R)(3+\Delta R)}$$

$$= \frac{12 * (12 + 4\Delta R - 12 - 8\Delta R - (\Delta R)^2)}{(6+\Delta R)(3+\Delta R)} \approx \frac{12 * (-4\Delta R)}{18} = \frac{8\Delta R}{3}$$

$$\therefore V_{th} = \frac{16}{3} * \left| \frac{\Delta R}{R} \right| \text{ Volts}$$

$$\therefore i = 20 \mu\text{A} = \frac{V_{th}}{R_{th} + 200} = \frac{\frac{16}{3} \left| \frac{\Delta R}{R} \right|}{2000 + 200}$$



$$\therefore \left| \frac{\Delta R}{R} \right| = 0.00825 = |\alpha \Delta T| \therefore |\Delta T| = \frac{0.00825}{0.01/^\circ\text{C}} = 0.825^\circ\text{C}$$

assumption is justified and new temperature =  $25 \pm 0.825 = 24.175^\circ\text{C}$  to  $25.825^\circ\text{C}$

$$\boxed{8-5} \text{ True power} = \frac{240^2}{20^2 + 18^2} * 20 = 1591.2 \text{ W} = 1.5912 \text{ kW}$$

$$\text{True pf} = \frac{20}{\sqrt{20^2 + 18^2}} = 0.74329$$

$$\textcircled{a} \text{ Wattmeter Reading} = 240 * \frac{240}{|21 + j19|} * \frac{21}{\sqrt{21^2 + 19^2}} = 1508.2 \text{ W} = 1.5082 \text{ kW}$$

$$\therefore \text{Reading} = 1.5082 \text{ kW} \text{ } \angle \text{ } \% \text{e} = 5.212 \%$$

$$\textcircled{b} \text{ pf reading} = \frac{21}{\sqrt{21^2 + 19^2}} = 0.74154$$

$$\therefore \text{Reading} = 0.74154 \text{ } \angle \text{ } \% \text{e} = 0.2366 \%$$

$$\boxed{8-6} \text{ a) Ideal power} = \left( \frac{240}{\sqrt{20^2 + 18^2}} \right)^2 * 20 = 1591.16 \text{ Watts}$$

Reading = power consumed by (load || voltage coil) =

$$\text{(load || voltage coil)} = (20 + j18) \parallel (1 + j1) \text{ K} = 26.40553 \angle 42.04^\circ$$

$$\therefore \text{Reading} = \left| \frac{240}{1 + j1 + 26.40553 \angle 42.04^\circ} \right|^2 * 26.40553 \cos 42.04^\circ$$

$$= 1459.633 \text{ watts}$$

$$\therefore \% \text{e} = 8.27 \%$$

$$\text{b) Ideal pf} = \frac{20}{\sqrt{20^2 + 18^2}} = 0.743294$$

$$\text{Reading} = \text{pf of (load || voltage coil)} = \cos 42.04^\circ = 0.742637$$

$$\therefore \% \text{e} = 0.0884 \% \approx \underline{\underline{-1\%}}$$

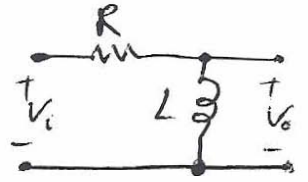
(68)

(7)

$$\boxed{9-2} \quad \frac{V_o}{V_i} = \frac{sL}{R+sL} = \frac{1}{1 + \left(\frac{R}{sL}\right)}$$

at dc:  $s=0 \quad \therefore \frac{V_o}{V_i} = 0$

at ac high frequency:  $s=\infty \quad \therefore \frac{V_o}{V_i} = 1$



$\therefore$  This is a high-pass filter.

$\therefore f_c = \infty \text{ Hz}$

$\neq f_c = \frac{\omega_c L}{2\pi}$  where  $\omega_c$  is given by  $\frac{R}{\omega_c L} = 1 \Rightarrow \omega_c = \frac{R}{L}$

$\therefore f_c = \frac{R}{2\pi L}$

for  $f = 2f_c = \frac{R}{\pi L} \quad \therefore \omega = 2\pi f = \frac{2R}{L} \quad \therefore s = \frac{j2R}{L} \quad \therefore sL = j2R$

$\therefore \frac{V_o}{V_i} = \frac{j2R}{R+j2R} = \frac{j2}{1+j2} = \frac{2}{\sqrt{5}} \angle 90^\circ - 63.4^\circ = \frac{2\sqrt{5}}{5} \angle 26.6^\circ = 0.894 \angle 26.6^\circ$

$\therefore$  Gain at  $f = 2f_c$  is 0.894

$$\boxed{9-3} \quad \frac{V_o}{V_i} = \frac{sL}{R+sL} = \frac{j\omega L}{R+j\omega L} \quad \text{at } \omega=0 \quad \therefore \frac{V_o}{V_i} = 0 \quad \text{at } \omega=\infty \quad \therefore \frac{V_o}{V_i} = 1$$

$\therefore$  It is a high pass filter  $\neq f_H = \infty \text{ Hz} \quad \neq f_c = \frac{R/L}{2\pi} = \frac{10/10m}{2\pi} = 159.2!$

$$\boxed{9-4} \quad \frac{V_o}{V_i} = \frac{R}{R+sL} = \frac{1}{1 + sL/R} = \frac{1}{1 + j\omega(10m/10)}$$

$$= \frac{1}{1 + j\omega m}$$

$\therefore \omega_m = 1 \quad \therefore \omega_c = \frac{1}{m} = 1 \text{ Krad/sec}$

$\therefore f_c = \frac{\omega_c}{2\pi} = 159.2 \text{ Hz}$

$\therefore$  The bandwidth is 159.2 Hz (69E)